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[AD-2]	Interpretation and Analysis of SKIM Campaign Observations (IASCO) (ESA Contract N° 4000129945/19/NL/FF/gp)	

### **Reference Documents**

Reference	Title
[RD-1]	DRIFT4SKIM 2018 Campaign Implementation Plan (ESA ref. 4000126110/18/NL/FF/D1)
[RD-2]	SKIM Campaign Data Acquisition Report (CNES ref. DSO/SI/TR-2019.01210)
[RD-3]	DRIFT4SKIM 2018 Data Acquisition Report (ESA ref. 4000126110/18/NL/FF/D2)
[RD-4]	Sutherland, P., SUMOS Campaign Summary, 2021
[RD-5]	Marié, L., F. Collard, F. Nouguier, L. Pineau-Guillou, D. Hauser, F. Boy, S. Méric, C. Peureux, G. Monnier, B. Chapron, A. Martin, P. Dubois, C. Donlon, T. Casal, F. Ardhuin: "Measuring ocean surface velocities with the KuROS and KaRADOC airborne near-nadir Doppler radars: a multi-scale analysis in preparation of the SKIM mission", <i>Ocean Sci.</i> , <b>16</b> , 1399–1429, 2020. https://doi.org/10.5194/os-16-1399-2020
[RD-6]	Earth Explorer 9 Candidate Mission SKIM –Report for Mission Selection, ESA-EOPSM-SKIM-RP-3550, v1.0, 21/06/2019.

### Signatures:

ESA: Craig Donlon

IFREMER : Bertrand Chapron,

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# On the Impact of Ocean Surface Motion on Delay-Doppler Altimetry

Louis Marié, Frédéric Nouguier, Doug C. Vandemark, Fabrice Ardhuin, and Bertrand Chapron

Abstract—The Poseidon-4 radar altimeter on board Sentinel-6 "Michael Freilich" (S6-MF) offers unique opportunities to more precisely estimate the impact of ocean surface motion on Delay-Doppler altimetry. The seminal "frozen-sea" analysis by [1] is extended to include the effect of these surface motions in the ensemble average Delay-Doppler Map signature of an isolated sea surface facet. Integrating this elementary signature over the instrument field of view, a fully analytical stacked echo waveform model is then derived, namely the IASCO echo waveform model Eq. (20). This IASCO waveform is validated, in the frozensea special case, with the well-established SAMOSA waveform model [2]. Sensitivities with respect to surface significant wave height, vertical velocity standard deviation and the "Geophysical Doppler" vector  $U_{GD}$  [3] projection along the satellite groundtrack velocity can then be directly evaluated. These developments provide theoretical and analytical means to jointly exploit S6-MF conventional and Delay-Doppler radar waveforms [4], to help refine estimates of Essential Climate Variables (Sea Level, Signficant Wave Height). With global upper ocean wave induced motions, systematic discrepancies between sea surface parameter estimates retrieved from S6-MF shall be mapped to possibly help reduce sea state biases on sea level instantaneous estimates.

Index Terms—Sentinel-6 MF, Delay-Doppler Altimetry, tracking waveform, surface displacement, surface motion.

### I. INTRODUCTION

A tellite radar altimetry provides unique data on sea level, surface wave significant height and surface wind speed. In practice, the determination of these parameters is obtained by fitting a theoretical function to the measured waveform, that is the observed averaged distribution of echoes as a function of time delay between the transmission and reception of radar pulses. This fitting is know as retracking and the actual methodology involves the choice of a theoretical waveform shape, and the choice of a cost function.

Improving the quality of the retrieved observations thus requires improving in a consistent way, not only the instrument characteristics, but also the accuracy of the parametric waveform model used in the retracking process. Before the instrument SIRAL on Cryosat-2, the information contained in the phase of the return signal from successive radar impulsions was discarded: only the echo signal intensity was considered, and averaged over bursts of impulsions to improve the Signal to Noise Ratio and mitigate speckle noise before retracking [5].

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D. Vandemark is with the Ocean Process Analysis Laboratory, University of New Hampshire, Durham, NH, USA.

Echo waveforms produced in this way were satisfactorily understood in the framework set by the original papers of [6] and [1], in which the overall waveform is considered to be the sum, on an intensity basis, of the echos returned by the individual scattering elements of the sea surface present in each range bin, assuming that the distribution of surface elevations is independent of the range bin [7]. As such, it can be modelled as the triple convolution [8] of the range "Point Target Response" (PTR), a function characterizing the intensity weighting in each range gate due to the instrument range resolution process, of the probability distribution function of the vertical displacement of scattering elements with respect to their rest position due to surface waves, and of a function describing the area of the sea surface encompassed in each range bin and the radiated power distribution due to the antenna system, the "Flat Surface Impulse Response" (FSIR).

Retrackers based on this waveform model typically retrieve three geophysical parameters from the radar signal [9]: the normalized radar cross-section  $\sigma_0$ , which is related to the surface slope statistics [10]–[12] the standard deviation of the vertical surface displacement, from which the significant wave height H<sub>s</sub> is derived, and the mean range from satellite to sea surface over the instrument footprint, which is later processed to give the Sea Level Anomaly (SLA), tides, etc.

The SIRAL instrument onboard CryoSat-2 was the first altimeter instrument to allow coherent (phase-preserving) processing of successive radar echos. The implementation of the Delay-Doppler Altimetry (DDA) concept proposed by [13], follows a Synthetic Aperture Radar (SAR) processing applied to successive echos to separate the signal contributions according to the along-track bins they return from. It results a two-dimensional Delay-Doppler Map (DDM) of the received power distribution across delay and Doppler. A "Range cell Migration Correction" (RCMC or RMC) is then applied to the waveforms from the different Doppler bins to realign them by removing the along-track distance contribution from the instrument-to-surface slant range. At this stage, signal accumulation ("stacking" or "multi-looking") is performed before a retracking of the resulting 1-D waveform is applied.

Several approaches have been proposed for the signal accumulation stage:

– In the "SAR altimetry" approach [14], SAR-processed radar waveforms viewing the same groundpoint during successive bursts of pulses are gathered together before summation. Fully-Focussed SAR, in which the final summation itself is phase-preserving, has even been successfully implemented [15], providing observations of the surface elevation of very compact water bodies: lakes, canals [16], sea-ice leads [17].

F. Ardhuin, B. Chapron, L. Marié and F. Nouguier are with the Laboratoire d'Océanographie Physique et Spatiale, Université de Brest, CNRS, Institut Francais de Recherche pour l'Exploitation de la Mer, IRD, 29280 Plouzané, France. e-mail: (Louis.Marie@Ifremer.fr).

– In the alternative "LR-RMC" approach recently proposed by [18], SAR-processing and RMC are applied on bursts of pulses, but are immediately followed by an incoherent stacking of the intensity. This approach does not yield a finer alongtrack resolution of the observations, but does provide a large increase in SNR for the subsequent retracking process.

Already clearly explained [13], the RMC has the effect of redistributing the total energy contained in the echo waveform, shifting the weight towards the instant of encounter of the radar pulse with the sea surface at nadir. The echo waveform is strongly affected in the process, featuring a sharp peak, instead of a step, at the instant of encounter. This redistribution of energy increases the SNR of the useful portion of the signal, thereby increasing the accuracy of the tracking process.

With the advent of DDA, and its subsequent development through the Sentinel-3A and Sentinel-6 "Michael Freilich" (S6-MF, [19]) missions, thus came the need to develop a new generation of retracking model waveforms, to serve as a basis for ground segment retrackers. Note, the two stacking strategies mentioned above yield equivalent waveforms when the situation is homogeneous in the along-track direction.

Starting from [20], devoted to the Cryosat-2/SIRAL waveform, a number of such DDA waveforms have been derived, with varying degrees of sophistication. [21] and [22] were the first to benefit from actual Cryosat-2 data. [22] proposes a very lucid exposition of the problem, analytically performs a number of steps, before resorting to the numerical evaluation of a number of integrals, with convincing comparisons with Cryosat-2 waveforms. [22], in the context of geodesy, goes further analytically, but overlooks the issue posed by the finite (and actually quite coarse) resolution of the along-track SAR processing. Had the method been used operationally, the proposed waveform peak is too narrow for given H<sub>s</sub>, leading to consistent overestimation of H<sub>s</sub>. In the framework of the ESAfunded SAMOSA project, [2] derived a semi-analytical model, since adopted by a large user community. Independently, [23] have proposed another semi-analytical model, putting specific emphasis on the retracking of waveforms suffering from instrument antenna mispointing.

These contributions were essentially developed within the original [1] framework, based on a triple convolution in the range direction only. Developments did not take into account the fact that the finite azimuthal resolution of the SAR processing and scattering facets motions (and not just height displacements) in effect also introduce a PTR and a probability distribution function in this direction. A more complete framework had to involve a triple convolution in the azimuth direction also.

Several steps in this direction were made by [4], [24]–[26]. These authors first proposed a computational approach involving a convolution with an azimuth PTR [24], then included a probability distribution function for the instantaneous radial velocity of the surface scattering elements [25], then discussed its statistics in relation with the overall sea state characteristics [26]. Finally, all these elements were consolidated in [4].

Though its initial motivation came from the conference presentation [27], the present contribution is derived independently of these works. Developments are performed within the framework of the ESA-funded "Interpretation and Analysis of SKIM Campaign Observations" (IASCO) project, focused on the analysis of airborne Doppler radar observations of the sea surface collected as part of the phase-A work for the Surface KInematics Multiscale (SKIM) concept that was a candidate mission for ESA Earth Explorer 9 [3], [28].

The structure of the article is the following: The problem context, including key instrument characteristics and acquisition geometry, is summarized in section II. A derivation of the instrument response to a single scattering facet, the basic building block of the overall instrument response, is discussed in section III. A very detailed consideration is given to the problem of statistical averaging over realizations of the random sea surface process, in terms of facet instantaneous elevation and velocity. The idea here is to reproduce for a moving sea surface the initial "frozen-sea" analysis of [1]. The probabilistic theory of these fluctuations is detailed in the gaussian sea state framework in Appendix A. The integration over the instrument Field of View (FoV), required to upscale the results obtained at the facet level to the altimeter waveform, is discussed in section IV. For computational convenience, the rationale followed is essentially that of the LR-RMC algorithm of [18]. However, as mentioned above, all the approaches give equivalent results in the along-track homogeneous situation in which model waveforms are usually derived. This section culminates with the analytical IASCO SAR altimeter waveform model Eq. (20), which is compared with the Open-Source pySAMOSA [29] implementation of the accepted SAMOSA model [2]. It is then extensively discussed, in section V. Conclusions and perspectives are finally presented in section VI.

### II. PROBLEM CONTEXT

### A. Main instrument characteristics

The main S6-MF orbit and Poseidon 4 instrument characteristics are summarized in table I. The mission is thoroughly described in [19]. Some of its main characteristics are briefly recalled here for the sake of completeness.

The mission main payload is Poseidon 4, a dual-frequency Ku-/C-band altimeter. A notable difference with respect to its predecessors of the Poseidon series of instruments is the fact that the chirp deramping process is performed numerically after digitization, instead of analogically in the receiver frontend. It is capable to work in the so-called "open-burst" mode, where it continuously alternates between transmission and reception of radar pulses. This is a significant improvement with respect to the SAR modes of the Cryosat-2 and Sentinel-3A instruments, which had to alternate between much longer burst transmission and burst reception phases ("closed-burst" mode). This entails in particular that the signal can be processed using both the CA and DDA approaches at all times. One downside of the open burst mode, however, is that the pulse travel-time sets strong constraints on the pulse repetition frequency  $f_p$ , which has to be roughly halved with respect to the previous instruments, and to vary continuously, to adjust for traveltime variations along the platform orbit. Finally, the instrument features a significantly improved onboard processor, which is capable of performing onboard the range-resolution, alongtrack Fourier transformation, and RMC steps of DDA processing. This capability is however not in permanent use, and the DDA processing is currently performed on ground.

Parameter	Symbol	Value
Mean flight altitude	h	$1347~\mathrm{km}$
Mean flight velocity	$v_s$	$6967 \text{ m.s}^{-1}$
Mean Earth Radius	$R_E$	$6371~\rm km$
Orbital factor	$\kappa = 1 + h/R_E$	1.21
Average pulse repetition frequency	$f_p$	$9178~\mathrm{Hz}$
Carrier frequency (Ku-band)	$f_c$	$13.575~\mathrm{GHz}$
Carrier wavelength (Ku-band)	$\lambda$	2.21 cm
Pulse duration	$T_p$	$32 \ \mu s$
Chirp bandwidth	В	$-320 \mathrm{~MHz}$
ADC sampling frequency	$f_s$	$395 \mathrm{~MHz}$
ADC I/Q number of samples	$N_s$	20480
Chirp rate	s	$-1.010^{13}~{\rm Hz.s^{-1}}$
Range-resolution effective time shift	$\delta t_{rr} = \frac{h_{trk}}{c} + \frac{f_c}{s}$	$3.13 \mathrm{~ms}$
Pulses per burst	$N_p$	64
Burst duration	$N_p/f_p$	$6.97 \mathrm{\ ms}$
Antenna 3 dB half-beamwidth	$\frac{\theta_{3dB}}{2}$	$0.665^{\circ}$
2-way ambiguity velocity	$\frac{\lambda f_p}{4}$	$\sim \pm 50.7~{\rm m.s^{-1}}$
2-way ambiguity $y$	$\frac{\lambda f_p}{4} \frac{h_0}{v_s}$	$\sim\pm9.8~{\rm km}$
1-way range diversity over FoV	$r - h_0 \simeq \kappa \frac{y^2}{2h_0}$	$\sim [0;\;43.2~{\rm m}]$
Elevation angle $\theta$ at end of unambiguous zone	$rac{\lambda f_p}{4 v_s}$	$0.417^{\circ}$

TABLE I

SENTINEL 6-MF / POSEIDON-4 DESIGN PARAMETERS.

One other consequence of the low  $f_p$  of Poseidon 4 is the fact that the Nyquist frequency  $f_p/2$  of the SAR processing is also quite low. A convenient prescription for the 1-way power radiation diagram G of the instrument antenna is a Gaussian dependence on the elevation angle  $\theta$  [1], [21]:

$$G(\theta) = G_0 \exp\left(-\frac{2\theta^2}{\gamma}\right), \ \gamma = \frac{\sin^2(\theta_{3dB})}{2\log(2)} \tag{1}$$

where  $G_0$  is the boresight gain of the instrument,  $\theta_{3dB}$  is the antenna half-power beamwidth. As can be seen in table I, the elevation angle at the along-track position where the Doppler frequency shift due to the platform flight velocity is equal to the Nyquist frequency is of the order of  $0.42^\circ$ , *i.e.* actually smaller than the antenna 3 dB half-beamwidth,  $0.66^\circ$ . A significant portion of the instrument Field of View (FoV) is thus aliased in the SAR processing, and appears in conspicuous sidelobes in the DDM.

### B. Measurement geometry

1) From the side: Figure 1 presents a side-view of the satellite trajectory and acquisition geometry, in the vertical



 $R_E$ 

ς

h

N

0

Fig. 1. Side-view of the nadir altimeter acquisition geometry. The Earth center, satellite antenna center-of-phase and nadir are respectively denoted O, S and N. The satellite is at altitude h above the reference spheroid of radius  $R_E$ . The satellite flight velocity with respect to the Earth surface is  $\vec{v}_s$ , with norm  $v_s$ , tangential component  $v_t$ , radial component  $v_r$  (positive upwards). At the current observation point P, the Earth surface is at height  $\eta$  above the reference spheroid of radius  $R_E$ , the elevation angle is denoted  $\theta$ , the satellite zenith angle is denoted  $\psi$ , the geocentric distance from nadir is  $\alpha$ . The vertical distance between the reference spheroid and its tangent plane at nadir is denoted  $\Delta$  (it is always negative). The vector from S to P is denoted  $\vec{r}$ , and its norm, the radar slant range to P, is denoted r.

 $\frac{\zeta}{\alpha}$ 

plane locally tangent to the trajectory in the Earth-bound frame [30]. A number of useful coordinate systems can be defined (see also Fig. 2). The current observation point P can be referred to by its geocentric distance from nadir  $\alpha$ , azimuth with respect to the satellite ground track  $\varphi$  and distance from the Earth center  $R_E + \eta$ , or by its cylindrical ( $\rho, \varphi, z = \Delta + \eta$ ) or Cartesian ( $x, y, z = \Delta + \eta$ ) coordinates in the frame of reference locally tangent to the sphere and centered on the nadir.  $\Delta$  denotes the distance in the z direction between the reference spheroid and the plane tangent to the sphere at nadir. It is always negative. In this work the y-direction of the Cartesian coordinate system is parallel to the satellite velocity vector.

Given the typical beam apertures used in nadir altimetry,  $\rho$  is of the order of at most 10<sup>4</sup> m. At such scales, it is easy to see that the observation point distances from nadir following the spherical surface,  $\zeta$ , and along the tangent plane ( $\rho$ , equal to  $\sqrt{x^2 + y^2}$ ), differ by less than 5 mm, and can be used interchangeably. From this one gets that  $\theta \simeq \tan(\theta) = \rho/h$ .

Clearly, however, the Earth sphericity must be taken into account for the vertical deflection of the surface, with  $\Delta$  varying with  $\rho$  as  $\Delta = -\rho^2/(2R_E)$  and reaching values of the order of 8 m at the edge of the instrument beam. This effect is also felt in the expression of the slant range from S to P,

$$r = \sqrt{\left(h - \eta + \frac{\rho^2}{2R_E}\right)^2 + \rho^2} \simeq h - \eta + \frac{\kappa}{2h}\rho^2,$$

which increases faster with  $\rho^2$  than in the flat-Earth approximation by a factor  $\kappa = 1 + h/R_E$ , equal to 1.21 for the Sentinel-6 MF flight parameters. In a similar way, though the elevation angle from the satellite,  $\theta$ , is well approximated by its flat-Earth expression  $\frac{\rho}{h}$ , the satellite zenith angle has to be corrected for sphericity and is given by



Fig. 2. Top-view of the nadir altimeter acquisition geometry. The Nadir is denoted by N. The y-direction of the Cartesian coordinate system is parallel to the ground projection of the satellite flight velocity vector,  $\vec{\mathbf{v}}_{s}$ . The azimuth angle  $\varphi$  is counted counterclockwise from the y-direction. The "Geophysical Doppler" vector is  $\vec{\mathbf{U}}_{GD}$ . The current point P can be referred to by its cylindrical coordinates  $(\rho, \varphi, \Delta + \eta)$ , its Cartesian coordinates  $(x, y, \Delta + \eta)$ , or by its Delay-Doppler coordinates  $r = \sqrt{(h - \Delta - \eta)^2 + \rho^2}$  and f. The isorange lines are represented by the circular grid centered on N, the isodops in absence of mean surface motion by the dashed lines grid, and the actual isodops showing the effect of mean surface motion by the thin lines grid.

2) From above: Figure 2 represents the acquisition geometry seen from above. The satellite flight velocity is directed along the positive y-direction. The current point P on the surface can be labeled using different sets of coordinates, Cartesian  $(x, y, z = \Delta + \eta)$ , cylindrical  $(\rho, \varphi, z = \Delta + \eta)$ , or, more interestingly, a Delay-Doppler coordinates set constructed from its slant range from the satellite, r, equal to  $h - \eta + \frac{\kappa}{2h}\rho^2$ , and the Doppler shift at its location, f.

At a given location, this Doppler shift is composed of a "frozen-sea" contribution and a wave-induced contribution emerging from the correlation of the Lagrangian velocity of the specular facets and their backscattered cross-sections. This contribution is discussed in section III and Appendix A, where it is shown that it can be represented as the projection on the radar line-of-sight of an additional "Geophysical Doppler" vector  $U_{GD}$ . In [3], this vector has been shown to be further decomposed in a "Current Doppler" vector  $U_{CD}$ , *i.e.* the ocean "Total Surface Current Vector" and a "Wave Doppler" vector due to the intrinsic motion of the sea surface specular facets. Note, this "Wave Doppler" vector is the dominant component (see Eq. 2 in [31]), with magnitude of the order of 2.5 to 3.5 m.s<sup>-1</sup> at Ku-band, and a weak dependence on sea state and wind speed. Denoting the magnitude of this vector,  $U_{GD}$ , and the azimuth it points to,  $\varphi_{GD}$ , the expression for the total Doppler frequency shift can be read from Eq. (13) as

$$f = \frac{2}{\lambda} \left[ -v_r + \frac{yv_t}{h} - \tan(\psi) U_{GD} \cos(\varphi - \varphi_{GD}) \right],$$

with  $\lambda$  the wavelength of the carrier radio waves. The radial velocity term is discussed in section III. It is easily removed as a "Doppler Centroid Anomaly", and does not warrant much discussion. Turning to the Geophysical Doppler contribution,  $U_{GD}$  is small with respect to  $v_t$ . Introducing  $\varepsilon$ , the fractional Doppler shift change observed due to the surface motion when looking along the satellite flight direction, equal to

$$\varepsilon = -\kappa \frac{U_{GD}}{v_t} \cos(\varphi_{GD}),$$

and the deflection angle

$$\varphi_{\varepsilon} = -\kappa \frac{U_{GD}}{v_t} \sin(\varphi_{GD}),$$

f can be expressed to first order in  $U_{GD}/v_t$  as

$$f = -\frac{2v_r}{\lambda} + \frac{2v_t}{\lambda h}\rho(1+\varepsilon)\cos(\varphi-\varphi_{\varepsilon}).$$
 (2)

When the satellite is going "against the wind",  $\varphi_{GD} \simeq 180^{\circ}$ , and  $\varepsilon$  is positive, indicating an excess Doppler shift with respect to the "frozen sea" situation. For the Sentinel-6 MF flight parameters,  $\varepsilon$  is of order  $4.10^{-4}$ , and  $\varphi_{\varepsilon}$  is of order  $0.03^{\circ}$ at most. The frozen-sea isodops are represented as dashed lines in figure 2, while isodops accounting for a (much exaggerated) Geophysical Doppler contribution are represented in thin lines.

### III. NEAR-NADIR DELAY/DOPPLER RADAR: AT THE FACET LEVEL

Near-nadir radar backscattering is well represented in the so-called Physical Optics, Kirchhoff, approximation. The electromagnetic field backscattered to the instrument from the sea surface is dominated by the weighed composition of individual contributions backscattered from the different stationary-phase (*i.e.* "specular") facets of the surface [32], [33]. The generic traits of the life histories of such specular points have been thoroughly studied in the gaussian sea surface context by [34]. In short, such "facets" appear and disappear at well-defined instants, and have well-defined trajectories in between. They thus constitute meaningful objects, and can be ascribed Lagrangian trajectories and velocities during their lifetime.

The phases of their contributions to the overall field are given by the two-way distances of the facets to the phase center of the instrument antenna, modulo the wavelength, and are thus essentially random, uniformly distributed and statistically independent. Upon taking products of the field with itself at the same or different instants or locations and performing ensemble averages, only terms involving the product of the contribution of one facet with its complex-conjugate can survive. Practically, this means that interferences between the different scattering elements of the scene do not need to be accounted for in the computation of the average total received power. These assumptions lead to Rayleigh statistics for the field, and to the fact that the average backscattered power is proportional to the illuminated surface even though the field mathematical expectation remains zero, in analogy with the phenomenology of the Wiener process. Two other consequences are that the power Point Target Response of the instrument is equal to the squared modulus of its amplitude Point Target Response, and that applying the Range Migration Correction step of DDA to complex amplitudes or detected power yields the same result for the average power Delay/Doppler Map (DDM).

This section contains the derivation of the Delay/Doppler power response of the instrument to a single facet, the basic building block from which an understanding of the DDM obtained from near-nadir observations of a whole moving sea surface can be constructed.

### A. Facet kinematics

Consider an isolated facet whose rest position is  $(x_0, y_0, z_0 = \Delta + \eta)$ . It is affected by surface wave motion, and at the middle of the observation interval it is located at  $(x_0 + \tilde{x}, y_0 + \tilde{y}, z_0 + \tilde{z})$  and moves with respect to the Earth with a Lagrangian velocity  $(\tilde{u}, \tilde{v}, \tilde{w})$ . Assume this Lagrangian velocity to be constant, and small with respect to both the satellite velocity and the speed of light. If we denote by  $\tau_n$  the instant of transmission of the n-th radar pulse with respect to the flight time of the pulse since its transmission, the distance from the radar to the facet evolves with time  $\tau_n + t'$  as:

$$r(\tau_n + t') = \begin{cases} (x_0 + \tilde{x} + \tilde{u}(\tau_n + t'))^2 \\ + (y_0 + \tilde{y} + (\tilde{v} - v_t)(\tau_n + t'))^2 \\ + (h - (z_0 + \tilde{z}) + (v_r - \tilde{w})(\tau_n + t'))^2 \end{cases}$$

In this expression, many terms are very small with respect to the carrier wavelength over the ranges of variation of both t'and  $\tau_n$ , and can be neglected. A robust approximate expression becomes

$$r(\tau_n + t') \simeq h - (\eta + \tilde{z}) + \frac{\kappa \rho_0^2}{2h} + \left[v_r - y_0 \frac{v_t}{h} - \tilde{w}\right] \tau_n + \left[v_r - y_0 \frac{v_t}{h}\right] t' + \frac{v_s^2 \tau_n^2}{2h}.$$

The last term in this expression can be relevant in the Fully-Focused SAR altimetry context, for which coherent processing is used for  $\tau_n$  as large as 1 s. On the shorter time scales relevant for DDA, it remains negligibly small. The slant range from the radar to the facet is thus obtained as a function of the fast and slow times as:

$$r(\tau_n + t') \simeq \tag{3}$$

$$\underbrace{h + \frac{\kappa \rho_0^2}{2h} - \eta - \tilde{z}}_{r_0} + \underbrace{\left[v_r - y_0 \frac{v_t}{h} - \tilde{w}\right] \tau_n}_{r'_0 \tau_n} + \underbrace{\left[v_r - y_0 \frac{v_t}{h}\right] t'}_{r_0 t'}.$$

The vertical component of the Lagrangian velocity of the facet manifests itself as a slow pulse-to-pulse phase drift. The other facet velocity components are much too small to be felt. The large dominating contributions due to the platform velocity, however, must be taken into account on both the slow (interpulse,  $\tau$ ) and the fast (intra-pulse, t') time scales.

### B. Radar signal generation and reception

Consider a burst of  $N_p$  radar pulses, numbered from 0 to  $N_p - 1$ , transmitted every  $1/f_p$  seconds from the satellite between  $-T_b/2$  and  $T_b/2$ . Pulse number n is transmitted at instant  $\tau_n$ . Time between transmission and reception of a pulse is measured with the "fast time" variable t'. For each pulse, transmission thus lasts from  $t' = -T_p/2$  to  $t' = T_p/2$ . The return signal is recorded during a time window that is centered around an estimate of the two-way pulse travel time that is provided by an onboard tracker,  $t'_{trk} = 2h_{trk}/c$ . This time delay can be considered constant over the burst duration, and the recording window is long enough to contain the entire useful length of the return signal<sup>1</sup>.

Between  $t' = -T_p/2$  and  $t' = T_p/2$ , the instantaneous transmitted frequency changes linearly from  $f_c - B/2$  to  $f_c + B/2$ . Introducing the chirp rate  $s = B/T_p$ , the transmitted signals thus read:

$$S_{TX}(t') = \mathbb{1}_{[-T_p/2;T_p/2]}(t')e^{j2\pi f_c t'}e^{j2\pi \frac{st'^2}{2}}.$$

At first order in  $\dot{r_0}/c$ , the signal received at time instant  $\tau_n + t'$  has left the instrument at time  $\tau_n + t'(1 - 2\dot{r_0}/c) - 2\frac{r_n}{c}(1 - \dot{r_0}/c)$ . Introducing a slowly varying complex amplitude  $a(\tau_n)$  to keep track of the effect on the field of the reflection on the facet, and up to the geometric factors required to take into account the antenna radiation pattern, spherical divergence and propagation losses, which can all be reinstated at the end of the calculation, the received signal is:

$$S_{RX}(\tau_n + t') = a(\tau_n)S_{TX}\left(t'\left(1 - 2\frac{\dot{r_0}}{c}\right) - 2\frac{r_n}{c}\left(1 - \frac{\dot{r_0}}{c}\right)\right). \quad (4)$$

Upon reception, the signal is amplified and analogically multiplied by  $e^{-j2\pi f_c t'}$ , prior to recording. During the recording window, the return signal is "de-ramped" by a replica of the transmitted chirp, either analogically prior to digitization or, as in the Poseidon-4 instrument, numerically after digitization. This replica is centered on  $t' = t'_{trk}$ , and is thus better described as a function of yet another time scale,  $t = t' - t'_{trk}$ . One thus finally ends up with a digitized recording of

$$S(\tau_n, t) = e^{-j2\pi \left(f_c t' + \frac{st^2}{2}\right)} S_{RX}(\tau_n + t'_{trk} + t).$$

<sup>1</sup>A number of delicate intermediate steps are required to make this technologically feasible, in order to accommodate digitizer bandwidth or memory constraints, such as introduction, and subsequent compensation of, a digitization window delay with respect to the pulse start, pulse interleaving control, PRI modulation to accommodate altitude variations, etc. These steps are very well controlled, and we will in the following assume that the artefacts they introduce have been perfectly compensated.

### C. Range compression

Using the expression of  $S_{RX}$  given in Eq. (4), and neglecting terms amounting to phase shifts smaller than  $10^{-2}$  cycles, the expression of the digitized signal segments is

$$S(\tau_n, t) = a(\tau_n) \mathbb{1}_{\left[-T_p/2; T_p/2\right]} \left(t + 2(h_{trk} - r_n)/c\right)$$
$$\times e^{j4\pi \left[s \frac{\left(r_n - h_{trk}\right)^2}{c^2} - \frac{r_n}{\lambda} - \frac{r_0}{c} \frac{h_{trk}}{\lambda}\right]}$$
$$\times e^{-j4\pi \frac{st}{c} \left[r_n + \frac{f_c r_0}{s} - h_{trk} (1 - \frac{r_0}{c})\right]}.$$

The term on the first line keeps track of the envelope of the pulse. The term on the second line is a phase term, constant over one pulse, but slowly changes from pulse to pulse. The final term describes the time dependence of the deramped echo on fast time. This time-dependence is used to obtain the response of the instrument to the facet as a function of observation range<sup>2</sup>, through a Fourier transform on t. Using the fact that the independent variable r is linked to the analysis frequency  $\omega$  by the relationship  $r = h_{trk} - \frac{c\omega}{4\pi s}$ , one obtains the radar echo amplitude of the facet as a function of r at time  $\tau_n$  as:

$$S(\tau_n, r) = a(\tau_n)e^{j4\pi \left[s\frac{(r_n - h_{trk})^2}{c^2} - \frac{r_n}{\lambda} - \frac{r_0}{c}\frac{h_{trk}}{\lambda}\right]}$$

$$\times \frac{1}{T_p} \int \mathbb{1}_{\left[-T_p/2; T_p/2\right]} \left(t + 2(h_{trk} - r_n)/c\right)$$

$$\times e^{j4\pi \frac{st}{c} \left[r - r_n - r_0\left(\frac{h_{trk}}{c} + \frac{f_c}{s}\right)\right]} dt.$$

$$= a(\tau_n)e^{j4\pi \left[s\frac{(r_n - h_{trk})^2}{c^2} - \frac{r_n}{\lambda} - \frac{r_0}{c}\frac{h_{trk}}{\lambda}\right]}$$

$$\times \Upsilon_r \left(r - r_n - \dot{r_0}\left(\frac{h_{trk}}{c} + \frac{f_c}{s}\right)\right).$$

In this expression  $\Upsilon_r(\delta r) = \operatorname{sinc}\left(\frac{2\pi sT_p}{c}\delta r\right)$  denotes the range-compression amplitude impulse-response function, which is symmetric and peaking around 0. At each instant, the range-resolved waveform is thus peaked close to the actual facet range  $r_n$ , but includes cm-scale corrections for the motion of the instrument during the travel time of the radar pulse ( $\dot{r_0}h_{trk}/c$  term) and for delay-Doppler ambiguity  $(\dot{r_0}f_c/s \text{ term})$ . The first of these two corrections can be absorbed in a change of the definition of  $\tau_n$ , from the instant of pulse transmission to the instant the pulse strikes the facet (this convention has been used by [2]). The latter correction is of a different nature, and has its origin in the range-resolution method itself: the only differences allowing to discriminate a time-shifted and a Doppler-shifted versions of a chirp occur at its ends, in a vanishingly small part of its total duration. A target observed from a moving instrument thus appears shifted from its actual location, by an amount that depends on the relative speed and on the chirp bandwidth and direction. The facet thus appears shifted by  $\dot{r_0}\delta t_{rr}$  in the range-resolved waveform, with  $\delta t_{rr} = \frac{h_{trk}}{c} + \frac{f_c}{s}$  an effective time shift, equal to 3.13 ms for the S6-MF flight parameters. The sensitivity of

this term on  $h_{trk}$  is small, and it is adequately evaluated using the nominal flight altitude h. From Eq. (3),  $\dot{r_0} = v_r - y_0 v_t / h$ . The part of this range correction, due to the radial velocity of the platform  $v_r$ , is uniform in y and identically affects the whole scene. More importantly, the part due to the tangential velocity  $v_t$ , however, depends on the along-track location of the facet, and can only be removed after this is estimated in the Doppler-resolution stage.

The peak itself has a cardinal-sine shape, which stems from the amplitude envelope of the chirp. The Fourier-space representation of its square, which will be needed in the following, is readily derived to be the "triangle" function

$$\widehat{\Upsilon_r^2}(\mathcal{K}) = \sigma_r \mathbb{1}_{[-1;1]} \left(\frac{\mathcal{K}\sigma_r}{2\pi}\right) \left[1 - \frac{|\mathcal{K}|\sigma_r}{2\pi}\right], \qquad (5)$$

with  $\sigma_r = \frac{c}{2B}$ . For the Sentinel 6-MF parameters, the half-power width of this function is equal to  $0.886 \times \frac{c}{2B} \simeq 0.415$  m. Classically [1], [6], it is approximated by a Gaussian dependence, though some confusion exists regarding the appropriate width. One choice is to select the value for which the -3 dB widths of the cardinal-shape shape and the Gaussian approximation coincide. For the power point target response (PTR) in the range direction, this yields

$$\Upsilon^2_r(\delta r) \simeq e^{-\frac{\delta r^2}{2\sigma_r^2}}$$

using this time  $\sigma_r = \frac{1}{\sqrt{2\log(2)}} \frac{0.886c}{4B} = 0.176$  m. The equivalent expression in Fourier space reads

$$\widehat{\Upsilon}_r^2(\mathcal{K}) \simeq \sqrt{2\pi} \sigma_r e^{-\frac{\mathcal{K}^2 \sigma_r^2}{2}}.$$
(6)

### D. Doppler resolution

The next step in the evaluation of the contribution of the individual facet to the DDM is to study the effect of the Doppler frequency resolution. This step is performed through a weighted Fourier transform in slow time. The discrete sampling in  $\tau$  tends to complicate notations. It can obscure those characteristics of the DDM that are due to geophysical processes with others that are due to technicalities (typically, aliasing in Doppler frequency). We will thus conduct the analysis as if the sampling was continuous in slow time. We discuss the changes introduced by discrete sampling when relevant. In this framework, the echo amplitude waveform due to the facet at slow time  $\tau$  can be modeled as

$$S(\tau, r) = a(\tau)e^{j4\pi \left[s\frac{(r_0 - h_{trk})^2}{c^2} - \frac{r_0}{\lambda} - \frac{h_{trk}}{\lambda}\frac{\dot{r_0}}{c}\right]} \times e^{-j4\pi \frac{\tau r_0'}{c} \left[f_c - 2s\frac{r_0 - h_{trk}}{c}\right]} \times \Upsilon_r \left(r - \left[r_0 + \tau r_0' + \dot{r_0}\delta t_{rr}\right]\right).$$

The phase term on the first line of this expression is constant even in slow time, and can in fact be subsumed in the complex return signal amplitude  $a(\tau)$ . The term on the last line encodes the fact that the return waveform is essentially zero, except in the close neighborhood of the actual facet range. Of the phase terms in the second line, it can be seen that only the carrier frequency term can induce a noticeable contribution. It is the

<sup>&</sup>lt;sup>2</sup>In the original article of [13], the range compression is performed after the along-track Fourier transform. The order of these two stages is unconsequential, and we have based our description on the order used in the S6-MF onboard processor.

Doppler shift term the DDA method is based on. The others can be neglected.

The amplitude signature of the facet in the Delay-Doppler plane<sup>3</sup> is obtained by performing a weighted Fourier transform in  $\tau$ :

$$S(f,r) = \frac{1}{T_b} \int w(\tau) a(\tau) e^{-j2\pi\tau \left(f + 2\frac{r_0'}{\lambda}\right)} \\ \times \Upsilon_r \left(r - \left[r_0 + \tau r_0' + \dot{r_0} \delta t_{rr}\right]\right) \mathrm{d}\tau.$$

In this expression,  $w(\tau)$  is a "window" weighting function, used to control the shape of the impulse response function along the f dimension.

A small  $\tau$ -varying term appears in the argument of the range-compression impulse response function. For a fixed value of r, this term amounts to a slowly-varying perturbation of the window function used in the Fourier transform, giving more weight to the beginning, center or end of the measurement interval depending on whether the facet is entering, dwelling in, or leaving the neighbourhood of r. The effect of this term is small, broadening slightly the result in the range direction for the large  $r'_0$  bins. Neglecting it (which amounts to using for  $\Upsilon_r$  its value at the middle of the observation interval), we obtain

$$S(f,r) = \Upsilon_r \left( r - [r_0 + \dot{r_0} \delta t_{rr}] \right) \\ \times \frac{1}{T_b} \int w(\tau) a(\tau) e^{-j2\pi\tau \left( f + 2\frac{r'_0}{\lambda} \right)} d\tau.$$

### E. Ensemble averaging over the sea surface realizations

Taking the squared modulus of S(f, r), the power response for this realization of the facet is:

$$|S|^{2}(f,r) = \Upsilon_{r}^{2} \left(r - [r_{0} + \dot{r_{0}}\delta t_{rr}]\right)$$
  
$$\frac{1}{T_{b}^{2}} \iint w(\tau)w(\tau')a(\tau)a^{*}(\tau')e^{j2\pi(\tau'-\tau)\left(f + 2\frac{r_{0}'}{\lambda}\right)} \mathrm{d}\tau \mathrm{d}\tau'.$$

The mathematical expectation of the DDM contribution per unit projected area of the facets located around  $(x_0, y_0, \Delta + \eta)$ must be obtained by averaging this individual contribution over the possible realizations of the sea surface.

Sources of randomness that must be considered are N, the actual number of facets, with probability  $p_N$ , dependent on the location and area of the patch, and for each facet, labelled n,

- its corresponding vertical displacement due to waves,  $\tilde{z}$ ,
- its corresponding vertical Lagrangian velocity,  $\tilde{w}$ ,
- its corresponding complex reflecting amplitude at times  $\tau$  and  $\tau'$ ,  $a(\tau)$  and  $a(\tau')$ .

Hence, in the most general setting, the overall DDM contribution of a unit area patch located in the neighbourhood of  $(x_0, y_0, \Delta + \eta)$  should be expressed as:

The probability distribution function  $\sum_{N} p_N \sum_{n}^{N} P_n(\cdots)$  is a very high-dimensional function, of which only specific dependencies are currently known:

- The number of facets N and their complex reflection coefficient a are not known separately, but [33]  $\sum_N p_N \sum_n^N \int |a|^2 P_n(|a|^2) d|a|^2$  is proportional to the normalized sea surface backscatter cross section,  $\sigma_0$ , whose statistics and dependencies with respect to environmental conditions and observation geometry have been extensively studied.
- The statistics of  $\tilde{z}$ , independently of the other variables, have also been studied extensively. A Gaussian prescription is a reasonable assumption, though higher-order contributions likely exist, e.g. altimetric Sea State Bias (SSB).
- Statistical correlations of  $\tilde{z}$  and  $\sigma_0$  have been studied, and are known to result in altimetric electromagnetic bias (EMB).
- The joint statistics of  $\tilde{z}$  and  $\tilde{w}$  have been studied theoretically [35]. In the Gaussian context, these variables are independent.
- On the contrary, even in the Gaussian context, w̃ is not independent of the local sea surface slope ∇z̃, which in turn is correlated with the backscatter cross-section. A correlation between surface backscatter and along-line-ofsight surface velocity projection is thus expected, giving rise to the "Geophysical Doppler" bias.
- The slow-time dynamics of the complex backscatter amplitude a is very hard to study experimentally, and thus remains elusive, despite its fundamental importance in determining the optimal time window for coherentradar observations of the sea surface. Choosing a long time window for the coherent processing is not beneficial if the lifetime of the individual facets is short, as the expected resolution improvement is not achieved, but the opportunity to obtain independent looks is wasted. For instance, the effective Doppler (hence along-track) point target response function can be computed under the assumptions of a Gaussian dependence on  $\tau - \tau'$  of the correlation function  $\langle a(\tau)a^*(\tau')\rangle$  and of a Gaussian window function. This exercise shows that it is the shortest time scale (facet coherence time or window function width) that sets the effective Doppler resolution.

<sup>&</sup>lt;sup>3</sup>This is an instance where the discrete sampling in  $\tau$  introduces a difference with the time-continuous analysis, by making the impulse-response periodic in f with ambiguity period  $f_p$ . While this ambiguity is unconsequential for the high- $f_p$  instruments of CryoSat-2 and Sentinel-3, for which the sidelobes correspond to very large along-track distances and are rejected by the antenna diagram, the factor of two reduction in  $f_p$  of the Poseidon-4 instrument lets very conspicuous sidelobes enter the DDM. The power content in those sidelobes significantly contributes to the stacked DDA power waveform. They can be accounted for analytically, as will be shown in section IV-D.

Given all these knowledge gaps, we pursue the analysis according to the following practical assumptions:

- We consider all facets to be independent. The sums over facet number N and index n reduce to multiplication by  $\langle N \rangle$ , the mathematical expectation of N.  $\langle N \rangle$  depends on the observation geometry.
- We consider the elevations of the facets  $\tilde{z}$  to be Gaussian and independent of all other variables (i.e. the study of SSB and EMB is left for future investigation).
- · We consider the correlation time of the complex reflecting amplitude a to be long enough with respect to the observation window duration, i.e. its dynamics can be neglected. This assumption is probably at least marginally correct over the duration of one burst (6.97 ms). With this assumption, the product of Fourier transforms over slow time simplifies into a Doppler resolution PTR,  $\Upsilon_f^2(\delta f) = \frac{1}{T_b^2} \left| \int_{-T_b/2}^{T_b/2} w(\tau) e^{-j2\pi\tau\delta f} \right|^2$ . As discussed in [2], if the so-called "Hamming" window is used for  $w(\tau)$ , this PTR can be extremely well approximated by the Gaussian shape

$$\Upsilon_f^2(\delta f) \simeq \left(\frac{25}{46}\right)^2 e^{-\frac{\delta f^2}{2\sigma_f^2}},$$

with  $\sigma_f = \frac{1}{\sqrt{2\log(2)}} \frac{1.293}{2T_b} \simeq 78.74$  Hz. • After [32], [33], the sea surface NRCS is expressed as  $\sigma_0 = |R|^2 \langle N \rangle \pi \langle 1/|\widetilde{\Omega}| \rangle$ , with  $|R|^2$  the normalincidence Fresnel power reflection coefficient and  $\Omega$  the Gaussian curvature at specular points. Our placeholder  $|a|^2$  can thus be linked to the physical and geometric characteristics of the facet by  $|a|^2 = |R|^2 \pi / |\widetilde{\Omega}|$ .

F. Averaging over  $\tilde{z}$ : recovering the "triple-convolution model"

With these assumptions, the expression of the DDM contribution per unit area of the neighbourhood of  $(x_0, y_0, \Delta + \eta)$ reads

$$\langle |S| \rangle^2 (f,r;x_0,y_0,\Delta+\eta) = \pi |R|^2 \langle N \rangle$$

$$\times \int P_{sp}(\tilde{z}) \Upsilon_r^2 \left( r + \tilde{z} - h + \eta - \frac{\kappa}{2h} \rho_0^2 - \dot{r_0} \delta t_{rr} \right) d\tilde{z}$$

$$\times \iint P_{sp}(\tilde{w},\tilde{\Omega};x_0,y_0) \Upsilon_f^2 \left( f - 2\frac{\tilde{w}}{\lambda} + 2\frac{\dot{r_0}}{\lambda} \right) \frac{d\tilde{w} d\tilde{\Omega}}{|\tilde{\Omega}|},$$
(8)

where  $\Upsilon_r^2$  and  $\Upsilon_f^2$  are approximately Gaussian, with specified widths, and  $P_{sp}(\tilde{z})$  and  $P_{sp}(\tilde{w}, \tilde{\Omega})$  respectively provide the (Gaussian) pdf of the elevation and joint pdf of the vertical velocity and the total curvature of the surface at specular *facets*. Since we have chosen to neglect  $\tilde{z}$ 's correlations with the other variables,  $P_{sp}(\tilde{z})$  is identical to the full-surface pdf of  $\tilde{z}$ ,  $P(\tilde{z})$ .

The first integral is thus the convolution of a Gaussian with the instrument range PTR:

$$\widetilde{\Upsilon}_{r}^{2}(\delta r) = \int P(\tilde{z})\Upsilon_{r}^{2}\left(\delta r + \tilde{z}\right) \mathrm{d}\tilde{z}.$$
(9)

This integral defines an effective range PTR, accounting for both the effect of surface waves and the instrument PTR (such effective quantities will be denoted by a  $\tilde{\cdot}$  symbol in the following). It is the "frozen-sea" contribution to the detected interface thickness. It is accounted for by the "triple convolution model" treatment of CA and DDA.

In the Gaussian sea state approximation,

$$P(\tilde{z}) = \frac{1}{\sqrt{2\pi\sigma_h}} e^{-\frac{\tilde{z}^2}{2\sigma_h^2}},$$

where the surface vertical displacement standard deviation  $\sigma_h$ is related to the significant wave height H<sub>s</sub> by

$$\sigma_h = \frac{\mathrm{H_s}}{4}.$$

Using the Gaussian approximation of the instrument PTR given by Eq. (6), the effective range PTR can thus be readily approximated in physical space as

$$\widetilde{\Upsilon_r^2}(\delta r) \simeq \frac{\sigma_r}{\widetilde{\sigma_r}} e^{-\frac{\delta r^2}{2\widetilde{\sigma_r}^2}},$$

with

$$\sigma_r = \frac{1}{\sqrt{2\log(2)}} \frac{0.886c}{4B}, \ \widetilde{\sigma_r} = \sqrt{\sigma_r^2 + \sigma_h^2}$$

A representation using the exact range PTR expression Eq. (5) can be obtained in Fourier space as

$$\widehat{\widetilde{Y}_r^2}(\mathcal{K}) = \sigma_r \mathbb{1}_{[-1;1]} \left(\frac{\mathcal{K}\sigma_r}{2\pi}\right) \left[1 - \frac{|\mathcal{K}|\sigma_r}{2\pi}\right] e^{-\frac{\mathcal{K}^2 \sigma_h^2}{2}},$$

where  $\sigma_r = \frac{c}{2B}$ . Except where explicitly mentioned, all graphs in this article are produced using this formulation.

We now turn to the second integral in Eq. (8), which involves the probability density function  $P_{sp}(\tilde{w}, \Omega)$ .

G. Averaging over  $\tilde{w}$ : the effect of wave motion on the Doppler PTR

As shown in [36], [37], this specular-points pdf follows from the full-surface pdfs as

$$P_{sp}(\tilde{w}, \tilde{\Omega}) = \frac{|\tilde{\Omega}| \left[ 1 + \partial_x \tilde{z} |_{sp}^2 + \partial_y \tilde{z} |_{sp}^2 \right]^2}{\langle N \rangle} \\ \times P(\partial_x \tilde{z} = \partial_x \tilde{z} |_{sp}, \partial_y \tilde{z} = \partial_y \tilde{z} |_{sp}, \tilde{w}, \tilde{\Omega}).$$

Under a Gaussian assumption, it shows that the curvature dependent terms in Eq. (8) in fact compensate: facets with a really small curvature have a very large cross-section, but are less frequent by an exactly compensating amount. At lowest order in  $\tilde{z}$ , the curvature  $\Omega$  is a product of second spatial derivatives of  $\tilde{z}$ . In a Gaussian framework, discussed in section 2.1 of [35], these derivatives are independent of both the wave slope and the vertical velocity.  $\Omega$  can then be marginalized out and integrated. The surface slope at specular facets is fixed by the geometry, and expressed in terms of the observation angles  $\psi$  and  $\varphi$ . There remains:

$$\langle |S| \rangle^2 (f, r; x_0, y_0, \Delta + \eta) = \frac{\pi |R|^2}{\cos^4(\psi)}$$
(10)  
  $\times \widetilde{\Upsilon}_r^2 \left( r - \left[ h - \eta + \frac{\kappa}{2h} \rho_0^2 + \dot{r}_0 \delta t_{rr} \right] \right)$   
  $\times \int P(\partial_x \tilde{z} = \partial_x \tilde{z}|_{sp}, \partial_y \tilde{z} = \partial_y \tilde{z}|_{sp}, \tilde{w})$   
  $\times \Upsilon_f^2 \left( f - 2\frac{\tilde{w}}{\lambda} + 2\frac{\dot{r}_0}{\lambda} \right) \mathrm{d}\tilde{w}.$ 

The pdf of  $\tilde{w}$  is discussed at length in appendix A, where its expression is shown to read

$$P(\tilde{w}, \psi, \varphi) = \frac{1}{\pi \sqrt{2\pi} \sigma_{\tilde{w}} \text{mss}_{\text{T}}} \exp\left(-\frac{\tan^2 \psi}{\text{mss}_{\text{T}}}\right) \\ \times \exp\left(-\frac{1}{2\sigma_{\tilde{w}}^2} \left[\tilde{w} + U_{GD} \tan(\psi) \cos(\varphi - \varphi_{GD})\right]^2\right),$$

with  $mss_T$  the total mean squared surface slope, and

$$U_{GD} = -M_{ss}^{-1} msv$$

the  $U_{GD}$  vector, with  $U_{GD}$  and  $\varphi_{GD}$  its magnitude and direction. The distribution for  $\tilde{w}$  is Gaussian, but not centered, with variance

$$\sigma_{\widetilde{w}}^2 = m_{tt} - \mathbf{msv}^T \ \mathbf{M_{ss}}^{-1} \ \mathbf{msv},$$

where  $m_{tt}$ ,  $\mathbf{M}_{ss}$  and  $\mathbf{msv}$  are defined in terms of spectral moments of  $\tilde{\eta}$  and its time and space derivatives. This variance is uniform on the sea surface, and is somewhat smaller than the full-surface variance of  $\tilde{w}$ , because part of the dispersion of  $\tilde{w}$  is absorbed in the variations of the projection of  $\mathbf{U}_{GD}$ on the radar line-of-sight. Introducing this expression in Eq. (10), one obtains:

$$\langle |S| \rangle^{2}(f,r;x_{0},y_{0},\Delta+\eta) = \frac{|R|^{2}}{\cos^{4}(\psi)} \frac{\exp\left(-\frac{\tan^{2}\psi}{\mathrm{mss}_{\mathrm{T}}}\right)}{\mathrm{mss}_{\mathrm{T}}} \quad (11)$$

$$\times \widetilde{\Upsilon}_{r}^{2}\left(r - \left[h - \eta + \frac{\kappa}{2h}\rho_{0}^{2} + \dot{r}_{0}\delta t_{rr}\right]\right)$$

$$\times \int \exp\left(-\frac{1}{2\sigma_{\widetilde{w}}^{2}}\left[\tilde{w} + U_{GD}\tan(\psi)\cos(\varphi - \varphi_{GD})\right]^{2}\right)$$

$$\times \frac{1}{\sqrt{2\pi}\sigma_{\widetilde{w}}}\Upsilon_{f}^{2}\left(f - 2\frac{\tilde{w}}{\lambda} + 2\frac{\dot{r}_{0}}{\lambda}\right)\mathrm{d}\tilde{w}.$$

With  $\Upsilon_f^2$  known to be Gaussian, the last integral is the convolution of two Gaussian functions. It yields the Doppler Impulse Response function, accounting for the spreading effect of surface wave motion on the relative velocity of the satellite and the facet:

$$\widetilde{\Upsilon}_{f}^{2}(\delta f) = \left(\frac{25}{46}\right)^{2} \frac{\sigma_{f}}{\widetilde{\sigma_{f}}} e^{-\frac{\delta f}{2\widetilde{\sigma_{f}}^{2}}}, \qquad (12)$$

with

$$\widetilde{\sigma_f}^2 = \sigma_f^2 + \frac{4\sigma_{\widetilde{w}}^2}{\lambda^2}.$$

Inserting this expression in Eq. (11), and introducing the instrument power radiation diagram G and the various factors required to account for the spherical divergence of the transmitted signal, the DDM contribution per unit transmitted power of the neighborhood of  $(x_0, y_0, \Delta + \eta)$  finally reads

$$dDDM(f,r;x_0,y_0,\Delta+\eta) = \frac{\lambda^2 G^2(\theta,\varphi)}{(4\pi)^3 r_0^4} \sigma_0(\psi,\varphi) dS \quad (13)$$
$$\times \widetilde{\Upsilon}_r^2 \left( r - \left[ h - \eta + \frac{\kappa}{2h} \rho_0^2 + v_r \delta t_{rr} - \frac{y_0 v_t}{h} \delta t_{rr} \right] \right)$$
$$\times \widetilde{\Upsilon}_f^2 \left( f + \frac{2}{\lambda} \left[ v_r - \frac{y_0 v_t}{h} + \tan(\psi) U_{GD} \cos(\varphi - \varphi_{GD}) \right] \right)$$

### H. Discussion

Figure 3 presents the DDM obtained for 7 individual patches of the sea surface located every 3000 m along the satellite track, not accounting for (top graphs) and accounting for (bottom graphs) the impact of surface waves on the signal.  $v_r$  has been set equal to 0 in this analysis.

A first striking feature is that the footprint of each patch is very localized in the DDM: each sea surface patch produces only a very small footprint over a burst duration, and even if averaging over 7 consecutive bursts was performed to produce the DDM ( $\frac{1}{20}$  s incoherent averaging time such as classicaly used), the signature of each patch would only describe a very short segment of the parabola. Note, the full parabolic signature, often shown in publications, results from the combination of many footprints of individual patches, and the full DDM must be estimated as an integral over the sea surface.

A second feature, mostly visible Fig. 3a and b, is that even in the "no-waves" case the frequency resolution achieved in a single burst is quite coarse in comparison with the total Doppler bandwidth (the ratio is equal to the number of inter-pulse intervals per burst, 63 in this case). The "Range Migration Correction" on which DDA processing is based, by applying a frequency-dependent range shift, leaks this Doppler spread into a large contribution to range spread. This can lead to a strong overestimation of the significant wave height by the parameter estimation algorithm at the re-tracking step.

The effective widths of the range and Doppler pointtarget responses  $\Upsilon_r^2$  and  $\Upsilon_f^2$ , estimated as a function of wind speed using the elevation spectrum model of [38] in its fully developed limit, are represented in figure 4. The wave contribution to the effective range PTR width rapidly increases with wind speed, and becomes dominant for wind speed larger than 6 m.s<sup>-1</sup>. This is consistent with the very conspicuous broadening of the PTR in the range direction, visible between figures 3b and 3d. On the opposite, the instrumental contribution to the effective Doppler PTR width dominates up to wind speeds in excess of  $13 \text{ m.s}^{-1}$ . The overall Doppler PTR width only weakly depends on wind speed, remaining between 80 Hz, its low-wind value, and 120-130 Hz at high wind speed. Again, this is consistent with the only marginal width increase of the PTR in the Doppler direction between figures 3b and 3d.

Going back to Eq. (13), the effect of a non-zero  $v_r$  on the DDM is minor: the two terms which involve  $v_r$  amount to shifts of the DDM, one in the range direction  $(v_r \delta t_{rr}$  term in the range PTR), and one in the Doppler direction  $(2v_r/\lambda$  term in the Doppler PTR, leading to a "Doppler Centroid" correction). While not negligible, these two effects can readily be corrected. In the following, we assume these corrections already perfectly applied, and omit the two terms.

More importantly, the white lines in Fig. 3 mark the locus of a flat ( $\eta = 0$ ) sea surface on the DDM under a number of approximations. The dotted lines represent the flat-Earth approximation, which is obviously far too coarse. The dash-dotted lines represent the spherical-Earth approximation, seemingly much better representing the observed behaviour, but still slightly off with respect to the center of the PTR in



Fig. 3. a) DDM contributions of 1 m<sup>2</sup> sea surface patches regularly spaced every 3000 m in the along-track direction between y = -9000 m and 9000 m, taking only into account the instrument point-target response. c) DDM contributions of 1 m<sup>2</sup> sea surface patches regularly spaced every 3000 m in the along-track direction between y = -9000 m and 9000 m, taking into account the smearing in the range and Doppler directions due to fully developed waves described by the spectral form [38], with a 12 m.s<sup>-1</sup> head wind (3.75 m significant wave height). b) and d): close-up views on the neighborhood of the y = 9000 m patch (red rectangles in subplots a) and c)). On all the graphs, the color shades represent the DDM, normalized by its maximal value, the dotted, dash-dotted, dashed and thick lines represent respectively the locus of the sea surface in the flat-Earth approximation, in the spherical-Earth approximation, account taken of the  $\delta t_{rr}$  term, and taking also into account the Geophysical Doppler term. The green line segments in suplots b) and d) represent the shift along the f direction of the patch in the interval separating one burst and the next and one burst and the sixth next (classical 20 Hz product). The thin white crosses in suplots b) and d) mark the tip of the Delay-Doppler PTR.



Fig. 4. Effective range  $(\tilde{\sigma_r}, \text{ blue}, \text{ left axis})$  and Doppler  $(\tilde{\sigma_f}, \text{ red}, \text{ right axis})$  widths of the wave-accounting Point Target Responses  $\widetilde{\Upsilon_r^2}$  and  $\widetilde{\Upsilon_f^2}$  as a function of wind speed, for the elevation spectrum model [38] at infinite fetch. Dashed lines mark the instrumental lower bounds, thin lines mark the waves-only contribution, and thick lines represent the total effective widths.

Fig. 3d. Adding the range-Doppler ambiguity correction  $\delta t_{rr} \dot{r_0}$  results in the dashed lines. While improved, reaching perfect agreement definitely requires to correct the U<sub>GD</sub> effects (solid line). This correction should be applied as a stretch in the f coordinate, but can be more conveniently expressed as a shift in the r coordinate for given f: taking  $x_0 = 0$  for simplicity, the contribution to the DDM observed at f is the one that

should appear at  $f/(1 - \kappa \cos \varphi_{GD} U_{GD}/v_t)$  or, using the notations of Eq. (2),  $f/(1 + \varepsilon)$ . Its spherical Earth deflection correction  $\Delta$  should thus not be evaluated as

$$\Delta(f)|_{U_{GD}=0} = \frac{\kappa}{2h} \frac{(fh\lambda)^2}{(2v_t)^2}$$

but as

$$\Delta(f)|_{U_{GD}} = \frac{\kappa}{2h} \frac{(fh\lambda)^2}{(2v_t)^2} \frac{1}{(1+\varepsilon)^2}$$

The effect of  $U_{GD}$  must thus be represented as a perturbation to the spherical Earth range deflection correction,

$$\delta \Delta(f) = \Delta(f)|_{U_{GD}} - \Delta(f)|_{U_{GD}=0}$$
  
=  $\Delta(f)|_{U_{GD}=0} \left[\frac{1}{(1+\varepsilon)^2} - 1\right]$   
 $\simeq -2\varepsilon \Delta(f)|_{U_{GD}=0}$ 

These corrections are represented as a function of the Doppler frequency in figure 5. The spherical Earth deflection correction (in blue) is clearly overwhelming. Yet, the other corrections, are far from being negligible in the very demanding context of precise nadir altimetry. The range-Doppler ambiguity correction ranges up to 15 cm at the ends of the unambiguous along-track zone, and the  $\delta\Delta$  correction accounting for the effect of the Geophysical Doppler contribution can reach several centimeters of magnitude in the 12 m.s<sup>-1</sup> situation (the discussion in Appendix A shows that this case is in fact representative of the frequently encountered ocean situations with  $|U_{10}| \geq 7 \text{ m.s}^{-1}$ ).

### IV. NEAR-NADIR DELAY/DOPPLER RADAR:



Fig. 5. Range corrections as a function of Doppler frequency for a sea surface patch located on the satellite ground-track. The blue line (left scale) represents the spherical-Earth range deflection. The solid red line (right scale, note the very different scale) represents the range-Doppler ambiguity correction  $\dot{r}_0 \delta t_{rr}$ . The dashed red line (right scale) represents the  $\delta\Delta$  correction for waves described by the spectral form [38], with a 12 m.s<sup>-1</sup> head wind and infinite fetch (3.75 m significant wave height).

### FROM DELAY/DOPPLER MAP TO ECHO WAVEFORM

Section III provided us with a thorough understanding of the instrument response to an infinitesimal ocean surface patch. The upscaling operation from facet response to complete DDM, and more specifically the ways in which it can be brought back to the "triple convolutional model", and the production of the stacked echo waveform, are addressed in this section.

### A. The Flat Surface Impulse Response

Equation (13) expresses the DDM contribution of an infinitesimal ocean surface patch located at  $(x_0, y_0, \Delta + \eta)$ . Considering the sea surface to be flat, particularizing to the  $v_r = 0$  case, assuming the antenna radiation diagram to be axisymmetric, the backscattering cross-section to be independent of  $\theta$  and  $\varphi$  in the very narrow solid angle actually illuminated by the radar, approximating the  $1/r_0^4$  term introduced by the spherical divergence by its value at nadir, integrating over the sea surface, denoting by primes the integration point coordinates, and finally introducing the expression (2) for the Doppler shift at the integration point, a simplified expression for the full DDM is:

$$\begin{split} \mathrm{DDM}(f,r) &= \frac{\lambda^2 \sigma_0}{(4\pi)^3 h^4} \iint_{Sea} G^2(\theta') \rho' \mathrm{d}\rho' \mathrm{d}\varphi' \\ &\times \widetilde{\Upsilon}_r^2 \left( r - h + \eta - \frac{\kappa}{2h} \rho'^2 + \rho' \cos(\varphi') \frac{v_t}{h} \delta t_{rr} \right) \\ &\times \widetilde{\Upsilon}_f^2 \left( f - \frac{2v_t}{\lambda h} \rho'(1+\varepsilon) \cos(\varphi'-\varphi_\varepsilon) \right), \end{split}$$

It is a 2-dimensional convolution of a combined range and Doppler PTR, accounting for the effect of waves both in terms of vertical displacement and Doppler broadening, with a Flat sea Surface Impulse Response (FSIR), accounting for the Geophysical Doppler bias,

$$\begin{aligned} \text{FSIR}(f,r) &= \frac{\lambda^2 \sigma_0}{(4\pi)^3 h^4} \iint_{Sea} G^2(\rho'/h) \\ &\times \delta_r \left( r - h + \eta - \frac{\kappa}{2h} \rho'^2 + \rho' \cos(\varphi') \frac{v_t}{h} \delta t_{rr} \right) \\ &\times \delta_f \left( f - \frac{2v_t}{\lambda h} \rho'(1+\varepsilon) \cos(\varphi'-\varphi_\varepsilon) \right) \rho' \mathrm{d}\rho' \mathrm{d}\varphi'. \end{aligned}$$

In this expression,  $\delta_r$  and  $\delta_f$  denote Dirac delta functions acting along the *r* and *f* directions.

At this point, a useful simplification is to replace the second integration coordinate, i.e. the azimuth  $\varphi'$ , by the Doppler frequency at the integration point  $\frac{2v_t}{\lambda h}\rho'(1+\varepsilon)\cos(\varphi'-\varphi_{\varepsilon})$ . Neglecting the very small misalignment angle  $\varphi_{\varepsilon}$  and an order  $\varepsilon$  correction in the range-Doppler ambiguity term, and introducing Eq. (1) for the antenna gain, one obtains a Doppler-distributed breakup of the FSIR as

$$FSIR(f, r) =$$

$$\frac{2\lambda^2 G_0^2 \sigma_0}{(4\pi)^3 h^4} \int_0^\infty \int_{-f_{max}}^{f_{max}} \delta_f \left(f - f'\right) \frac{e^{-4\frac{\rho'^2}{h^2 \gamma}}}{\sqrt{f_{max}^2 - f'^2}} \\ \times \delta_r \left(r - h + \eta - \frac{\kappa}{2h}\rho'^2 + \frac{\lambda}{2}f'\delta t_{rr}\right) \rho' \mathrm{d}\rho' \mathrm{d}f',$$

with  $f_{max}(\rho') = \frac{2v_t}{\lambda h}(1+\varepsilon)\rho'$ . Defining

$$\mu_{\varepsilon} = \frac{\kappa h \lambda^2}{8v_t^2 (1+\varepsilon)^2}, \ \nu = \frac{8}{\gamma \kappa h}, \tag{14}$$
$$\mathcal{A}_{\varepsilon} = \frac{2\lambda^2 G_0^2 \sigma_0 \sqrt{\mu_{\varepsilon}}}{(4\pi)^3 \kappa h^3},$$

and the frequency and range coordinates of the apex of the FSIR in the range/Doppler plane

$$f_A = \frac{\lambda \delta t_{rr}}{4\mu_{\varepsilon}}, r_A = h - \eta - \mu_{\varepsilon} f_A^2$$

the FSIR becomes

$$FSIR(f,r) = \mathcal{A}_{\varepsilon} \frac{H(r - r_A - \mu_{\varepsilon}(f - f_A)^2)}{\sqrt{r - r_A - \mu_{\varepsilon}(f - f_A)^2}} e^{-\nu[r - r_A]}$$
(15)

with  $H(\cdot)$  the Heaviside step function. Note, a very small correction in the antenna diagram term has been neglected. The effect of a non-zero  $\varepsilon$  in the expressions of  $f_A$  and  $r_A$  is also absolutely negligible. This FSIR expression is suitable for the study, in the triple convolution model framework, of the radar altimeter echo waveform of a "non-frozen" sea.

With the S6-MF parameters,  $f_A = 8.447$  Hz, and  $\mu_{\varepsilon} f_A^2 = 0.146$  mm. Though not completely negligible, these shifts can still be considered small and, being of technological origin, are deterministic to be corrected for. In the following, we consider them to be perfectly corrected ( $r_A = h - \eta$ ,  $f_A = 0$ ). This amounts to considering that the correction represented as a thick red line in Fig. 5 has been applied.



Fig. 6. a) Graph of the Flat Surface Impulse Response Eq. (15) as a function of Doppler frequency shift f and range  $r - r_A$ . FSIR replicas shifted by  $+f_p$  and  $-f_p$  have been added to represent the along-track sidelobes. b) Graph of the physical space DDM obtained by inverse-transforming the Fourier space expression Eq. (17) with  $\tilde{\sigma_f}^2 = 0$ . This DDM takes into account the wave-induced broadening of the PTR in the range direction only, for 3.75 m significant wave height. Fourier-space DDM replicas shifted by  $+f_p$  and  $-f_p$  have been added before the inverse transform to represent the along-track sidelobes. c) Graph of the physical space DDM obtained by inverse-transforming the Fourier space expression Eq. (17). This DDM takes into account the wave-induced broadening of the PTR in the range direction only, for 3.75 m significant wave height. Fourier-space DDM obtained by inverse-transforming the Fourier space expression Eq. (17). This DDM takes into account the wave-induced broadening of the PTR in the range and Doppler directions due to fully developed waves described by the spectral form [38], with a  $12 \text{ m.s}^{-1}$  head wind (3.75 m significant wave height). Fourier-space DDM replicas shifted by  $+f_p$  and  $-f_p$  have been added before the inverse transform to represent the along-track sidelobes. d) Graph of the Range Migration Corrected physical space DDM obtained by inverse-transforming the Fourier space expression Eq. (22), in the same geophysical conditions. Fourier-space DDM replicas shifted by  $+f_p$  and  $-f_p$  have been added before the RMC and inverse transform to represent the along-track sidelobes.

B. From the Flat Surface Impulse Response to the echo waveform

Going from the expressions of the Range/Doppler PTR and the FSIR to an expression for the DDA echo waveform requires the following steps:

- Computation of the Fourier transforms of the PTR and the FSIR along the range and Doppler dimensions.
- Multiplication of the Fourier transforms.
- Inverse Fourier transformation in the Doppler dimension.
- Multiplication by a Doppler-dependent mask to implement the Range Migration Correction.
- Integration in the Doppler dimension to perform the stacking operation.
- Inverse Fourier transformation in the range direction.

All but the very last of these steps can be analytically performed.

The first step is detailed in Appendices B-A and B-B. The Fourier transforms of the FSIR and the PTR are obtained as functions of two new independent variables  $\mathcal{T}$  (conjugate to f) and  $\mathcal{K}$  (conjugate to r) as

$$\widehat{\widehat{\mathrm{FSIR}}}(\mathcal{T},\mathcal{K}) = \mathcal{A}_0 \frac{\pi e^{-i\mathcal{K}r_A} e^{-\frac{\mathcal{T}^2}{4\mu_\varepsilon(\nu+i\mathcal{K})}}}{\sqrt{\mu_0}(\nu+i\mathcal{K})}$$

and

$$\widehat{\widetilde{\Upsilon_f^2}\widetilde{\Upsilon_r^2}}(\mathcal{T},\mathcal{K}) = \sqrt{2\pi}\sigma_f \left(\frac{25}{46}\right)^2 e^{-\frac{\mathcal{T}^2 \widetilde{\sigma_f}^2 + \mathcal{K}^2 \sigma_h^2}{2}} \widehat{\Upsilon_r^2}(\mathcal{K}),$$

with  $\widehat{\Upsilon}_r^2(\mathcal{K})$  either of the two Fourier-space representations Eq. (5) or Eq. (6) of the range PTR. Their product is:

$$\widehat{\widehat{\text{DDM}}}(\mathcal{T},\mathcal{K}) = \mathcal{A}_0 \frac{\pi \sqrt{2\pi} \sigma_f}{\sqrt{\mu_0}} \frac{25^2}{46^2} \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}}}{\nu + i\mathcal{K}} \widehat{\Upsilon_r^2}(\mathcal{K}) \\ \times e^{-\frac{\mathcal{T}^2}{2} \left[\widetilde{\sigma_f}^2 + \frac{1}{2\mu_{\varepsilon}(\nu + i\mathcal{K})}\right]}.$$
(16)

As a sanity check, this expression can be used to derive the CA echo waveform. This calculation, presented in Appendix B-C, indeed yields the classical expected results.

Noting this expression to be Gaussian in  $\mathcal{T}$ , its inverse Fourier transform in this variable is

$$\widehat{\text{DDM}}(f,\mathcal{K}) = \mathcal{A}_{\varepsilon}\sqrt{2}\pi\sigma_{f}\frac{25^{2}}{46^{2}}\frac{e^{-i\mathcal{K}r_{A}-\frac{\mathcal{K}^{2}\sigma_{h}^{2}}{2}}}{\sqrt{\nu+i\mathcal{K}}}\widehat{\Upsilon}_{r}^{2}(\mathcal{K})$$
$$\times \frac{e^{-\frac{\mu_{\varepsilon}(\nu+i\mathcal{K})f^{2}}{1+2\mu_{\varepsilon}(\nu+i\mathcal{K})\overline{\sigma_{f}}^{2}}}}{\sqrt{1+2\mu_{\varepsilon}(\nu+i\mathcal{K})\overline{\sigma_{f}}^{2}}}.$$
 (17)

The next processing stage is to multiply this function by a Doppler-dependent mask to apply the Range Migration Correction. The aim of this operation is to phase-shift the contributions at different f in order to ensure a constructive summation in the stacking stage. Based on the original treatment by [13], the phase correction is  $\exp(i\mu_0 \mathcal{K} f^2)$ . This step is discussed in Appendix B-D, and yields the DDA echo waveform in Fourier space as:

$$\widehat{W_{DDA}}(\mathcal{K}) = \mathcal{A}_0 \frac{\sqrt{2\pi}}{\sqrt{\mu_0}} \sigma_f \frac{25^2}{46^2} \\ \times \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}} \widehat{\Upsilon_r^2}(\mathcal{K})}{\sqrt{\nu + i\mathcal{K}} \sqrt{\nu - 2i\mathcal{K}(\varepsilon + \mu_0 \nu \widetilde{\sigma_f}^2) + 2\mu_0 \mathcal{K}^2 \widetilde{\sigma_f}^2}}.$$
 (18)

Analytically inverting the Fourier transform to obtain the echo waveform as a function of r is only straightforward in the  $\varepsilon = 0$ ,  $\widetilde{\sigma_f} = 0$  case, and using the Gaussian approximation of the range PTR. This is done in Appendix B-E, where the result by [22] is recovered. The general case is more difficult.

### C. Discussion

Figure 6a represents the FSIR in the (f, r) plane, account taken of the along-track sidelobes in the Doppler dimension (see section IV-D below). Restricting the attention to the unambiguous region of the plane, the locus of the FSIR displays the expected parabolic shape. It is slightly displaced due to the  $U_{GD}$  effect, but this effect is too faint to be measurable in this representation. Taking slices for fixed f (not shown), it varies from zero for  $r - r_A \le \mu_{\varepsilon}(f - f_A)^2$ , and follows the expected  $1/\sqrt{r - r_A - \mu_{\varepsilon}(f - f_A)^2}$  singularity on crossing the locus. At the ends of the interval, the  $f > f_p/2$ (resp.  $f < -f_p/2$ ) branch of the parabola is aliased into the  $f \ge -f_p/2$  (resp.  $f \le f_p/2$ ) part of the domain. Close to the ends of the interval, this induces a twofold increase of the recorded power density.

Figure 6b now represents the DDM in the (f, r) plane, in the  $\widetilde{\sigma_f} = 0$  Hz approximation, for 12 m.s<sup>-1</sup> head wind (3.75 m significant wave height). This plot thus includes the effect of the instrument PTR and waves in the range direction only. It corresponds to the conditions studied by [22]. As expected, the convolution by the wave-thickened instrument PTR introduces a strong smearing of the FSIR in the range direction.

Figure 6c on the opposite represents the DDM in the (f, r) plane for 12 m.s<sup>-1</sup> head wind (3.75 m significant wave height), taking account of the instrument and waves PTR in the range and Doppler directions. The Doppler component of the PTR smears the DDM in the Doppler direction and strongly increases the thickness of the DDM. This effect is particularly noticeable for large values of f, where the FSIR locus is more slanted with respect to the iso-range lines, leading to a stronger impact of the smearing in the f direction.

Finally, Figure 6d represents the range-migration corrected DDM in the (f, r) plane, in the same environmental conditions. The unambiguous part of the DDM has successfully been flattened to the  $r \simeq r_A$  portion of the domain. In this portion of the DDM, the smearing in f has been converted by the RMC into a strong excess of smearing in the r direction, which can easily be confused with a strong excess of significant wave height.

The sidelobes, on the other hand, have been straightened (the parabolic part of the RMC applies for every f), but not folded to  $r \simeq r_A$ . Their power density is thus missing from the  $r \simeq r_A$  region, and contaminating the  $r > r_A$  part of the domain, thus affecting the overall shape of the echo waveform. It turns out this issue can analytically be modeled. This is the subject of the next section.

### D. Accounting for the Doppler sidelobes of the Poseidon-4 instrument

The open-burst mode of acquisition of the Poseidon-4 instrument sets quite stringent constraints on the radar  $f_p$ : it has to vary to accommodate changes in flight altitude along the orbit, and has to remain in the vicinity of the quite low value of 9 kHz. Consequently, signals scattered from distant up/down track areas can experience phase shifts of more than a half period between consecutive pulses. This introduces sidelobes in the DDM: a bin located at (r, f) collects power that should appear at  $(r, f+f_p)$  and  $(r, f-f_p)$ . Higher order ambiguities are rejected by the instrument radiation diagram and play a negligible role.

These sidelobes are unconsequential in the CA context, as the applied processing is not dependent on f. In the DDA context, however, all echoes detected in the Doppler bin at f are range-shifted by  $\mu_0 f^2$ , including those which should have been shifted by  $\mu_0 (f + f_p)^2$  or by  $\mu_0 (f - f_p)^2$ . The total stacked echo waveform is thus better reproduced by stacking  $\widehat{DDM}(f, \mathcal{K})e^{i\mu_0\mathcal{K}f^2}$  for  $f \in \left[-\frac{f_p}{2}; \frac{f_p}{2}\right]$ ,  $\widehat{DDM}(f, \mathcal{K})e^{i\mu_0\mathcal{K}(f-f_p)^2}$  for  $f \in \left[\frac{f_p}{2}; \frac{3f_p}{2}\right]$  and  $\widehat{DDM}(f, \mathcal{K})e^{i\mu_0\mathcal{K}(f+f_p)^2}$  for  $f \in \left[-\frac{3f_p}{2}; -\frac{f_p}{2}\right]$ . The expression for the full DDM including the sidelobes is given in Appendix B-F Eq. (23).

After DCA correction, the DDM is even in f to a very good approximation. The contributions from the two sidelobes are thus identical. The sidelobes-accounting Fourier space DDA waveform can thus be expressed as

$$\widehat{W_{DDA}^{SL}}(\mathcal{K}) = 2 \int_{0}^{\frac{f_p}{2}} e^{i\mu_0 \mathcal{K}f^2} \widehat{\text{DDM}}(f, \mathcal{K}) \mathrm{d}f \qquad (19)$$
$$+ 2 \int_{\frac{f_p}{2}}^{\frac{3f_p}{2}} e^{i\mu_0 \mathcal{K}(f-f_p)^2} \widehat{\text{DDM}}(f, \mathcal{K}) \mathrm{d}f.$$

The calculation is detailed in appendix B-G. Finally, the Fourier-space sidelobes-accounting DDA waveform is obtained as

$$\widehat{W_{DDA}^{SL}}(\mathcal{K}) = \mathcal{A}_0 \frac{\sqrt{2\pi^{3/2}}}{\sqrt{\mu_0}} \sigma_f \frac{25^2}{46^2}$$
(20)  
 
$$\times \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}} \widehat{\Upsilon}_r^2(\mathcal{K})}{\sqrt{\nu + i\mathcal{K}} \sqrt{\nu - 2i\mathcal{K}(\varepsilon + \mu_0 \nu \widetilde{\sigma_f}^2) + 2\mu_0 \mathcal{K}^2 \widetilde{\sigma_f}^2}}$$
$$\times \left[ \operatorname{erf}\left(\frac{f_p \ \Xi_{\varepsilon}}{2}\right) + e^{-f_p^2 \frac{\Xi_{\varepsilon}^2}{4}} \operatorname{erfc}\left(f_p \left[\frac{\Xi_{\varepsilon}}{2} + \frac{i\mu_0 \mathcal{K}}{\Xi_{\varepsilon}}\right]\right) - e^{-f_p^2 \left[i2\mathcal{K}\mu_0 + \frac{9\Xi_{\varepsilon}^2}{4}\right]} \operatorname{erfc}\left(f_p \left[\frac{3\Xi_{\varepsilon}}{2} + \frac{i\mu_0 \mathcal{K}}{\Xi_{\varepsilon}}\right]\right) \right].$$

with  $\overline{\operatorname{erfc}}(x) = e^{x^2} \operatorname{erfc}(x)$  the exponentially scaled complementary error function,

$$\Xi_{\varepsilon} \simeq \sqrt{\mu_{\varepsilon} \frac{\nu - 2i\mathcal{K}(\varepsilon + \mu_{0}\nu\widetilde{\sigma_{f}}^{2}) + 2\mu_{0}\mathcal{K}^{2}\widetilde{\sigma_{f}}^{2}}{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma_{f}}^{2}}}$$

and we recall for the sake of compactness that

$$\begin{split} \mathcal{A}_0 \frac{\sqrt{2\pi^{3/2}}}{\sqrt{\mu_0}} &= \frac{\lambda^2 G_0^2 \sigma_0}{(8\pi)^{3/2} \kappa h^3}, \\ \varepsilon &= -\kappa \frac{U_{GD}}{v_t} \cos(\varphi_{GD}), \\ \mu_\varepsilon &= \frac{\kappa h \lambda^2}{8 v_t^2 (1+\varepsilon)^2}, \ \nu &= \frac{8}{\gamma \kappa h}, \\ \delta t_{rr} &= \frac{h}{c} + \frac{f_c}{s}, \ f_A &= \frac{\lambda \delta t_{rr}}{4 \mu_\varepsilon}, \\ r_A &= h - \eta - \mu_\varepsilon f_A^2, \\ \sigma_r &= \frac{c}{2B}, \ \sigma_f &= \frac{1}{\sqrt{2 \log(2)}} \frac{1.293}{2T_b}, \\ \widetilde{\sigma_f}^2 &= \sigma_f^2 + \frac{4\sigma_{\widetilde{w}}^2}{\lambda^2}, \\ \widetilde{\Upsilon_r^2}(\mathcal{K}) &= \sigma_r \mathbbm{1}_{[-1;1]} \left(\frac{\mathcal{K}\sigma_r}{2\pi}\right) \left[1 - \frac{|\mathcal{K}|\sigma_r}{2\pi}\right]. \end{split}$$

This analytically derived final expression is the IASCO DDA waveform model.

### V. DISCUSSION

### A. Comparison between different waveform models

Figure 7 presents waveforms obtained using the different expressions developed in the text, for a Hs = 3.75 m sea surface generated using the spectrum of [38] for  $12 \text{ m.s}^{-1}$  wind and infinite fetch. All the waveforms are generated using the same value of  $\mathcal{A}_0$  for normalization. They contain identical integrated energy, except for the dashed blue and dashed magenta curves, which only account for the returns from the unambiguous Doppler frequency band. Aside from this particular case, the curves only differ by the expected distribution of the received energy as a function of time.

The cyan curve, which represents the expected waveform from CA processing, has a very different shape from the others: after the sharp initial rise at the encounter of the radar pulse with the surface, the power return in all range bins is expected to be essentially identical, except for the slow decrease due to the antenna radiation diagram.

On the contrary, DDA processing brings a much larger portion of the received radar energy into the initial peak. All DDA model waveforms reflect this general behaviour. Significant differences encode the different sets of hypotheses under which these waveforms are derived.

Typically, the waveform derived by [22], accounting for the spreading effect in range of the vertical distribution of backscattering facets, but not accounting for the quite coarse frequency resolution of the along-track Fourier transform and the broadening due to the vertical orbital velocities (red curve), displays a much sharper initial rise. Its initial peak is much higher than the other DDA waveforms. Accounting for the



Fig. 7. Graph of altimeter waveforms computed for a 12m.s<sup>-1</sup> side wind, corresponding to  $Hs = 3.75 \text{ m}, \varepsilon = 0$ , for the S6-MF flight parameters and in a number of different processing configurations. The cyan curve represents the CA waveform Eq. (21). The red curve represents the  $\tilde{\sigma} = 0$  DDA waveform discussed in Appendix B-E, under the assumption of very large  $f_p$ . The green curve represents the DDA waveform Eq. (18), which takes into account the Doppler PTR width, but still assumes a very large  $f_p$ . The dashed blue curve represents the unambiguous Doppler frequency range  $[-f_p/2; f_p/2]$ contribution to the IASCO waveform model Eq. (20), and the continuous blue curve represents the full IASCO waveform model Eq. (20), including the effect of Doppler sidelobes. Finally, the magenta curves represent the multilook waveforms produced by the "pySAMOSA" [29] implementation of the established SAMOSA model [2] over the main lobe only (dashed magenta curve), and after modification to account for the sidelobe aliasing (continuous magenta curve). All the analytical curves correspond to the same  $A_0$  factor, chosen such that the  $\tilde{\sigma} = 0$  waveform is one at the epoch. The SAMOSA waveforms have been normalized to correspond to the same integrated energy as the blue waveforms.

additional smearing of the energy in the Doppler dimension leads to the quite different waveform depicted by the green line. The main peak is largely reduced, and a significantly earlier rise in energy (the "toe" mentioned in [22]).

This effect of the finite Doppler resolution of the alongtrack Fourier transform is taken into account by the SAMOSA model [2] of multi-look SAR altimeter returns, as well as by the earlier semi-analytical models [20], [21], [23]. For these model waveforms, a correction serendipitously emerges from the numerical integrations performed over azimuth. Yet, the broadening effect of the scattering facets motion is not taken into account by any of these models.

Figure 6 clearly shows that the halved  $f_p$  value of the S6-MF instrument with respect to the CryoSAT and Sentinel-3A instruments introduces aliasing in the along-track Fourier transform. Consequently, only the energy returning from the strip of the sea surface with a Doppler frequency shift comprised between  $-f_p/2$  and  $+f_p/2$  is correctly redistributed by the RMC step. The waveform obtained by stacking over this reduced frequency band is represented as a blue dashed line in Figure 7. The height of the main peak is reduced by almost 20% with respect to the green curve, obtained by extending the integration to  $\pm\infty$ . Taking into account the energy contained in the sidelobes (continuous blue curve) only marginally increases the height of the main peak, but slows the decrease of the signal tail. We stress that the total energy content of this waveform is identical to that of the very different-looking CA (cyan) or  $\tilde{\sigma}_f = 0$  (red) waveforms.

Waveforms produced using the well-established SAMOSA model [2] are depicted by magenta lines, either multi-looking over the unambiguous Doppler frequency range (dashed line) or, after *ad hoc* modification to take into account the aliasing effect, over the full frequency range (continuous line). In both cases, the SAMOSA waveforms feature a higher main peak and a somewhat reduced "toe" compared to their counterparts produced using Eq. (20) (blue lines). The waveforms rapidly match after to exhibit very similar shapes over the decay region. Overall, the IASCO waveform model Eq. (20) is in good agreement with the SAMOSA ones over the late peak and decay region. Differences in peak shapes are however significant, and will be discussed in the next section.

### B. Effect of the Gaussian approximation of the range PTR



Fig. 8. Graph of altimeter waveforms computed for a  $12\text{m.s}^{-1}$  side wind, corresponding to Hs = 3.75 m,  $\varepsilon = 0$ , for the S6-MF flight parameters. The continuous blue curve represents the waveform obtained using the full DDA waveform Eq. (20), considering for the instrument range PTR the exact squared cardinal sine shape Eq. (5). The continuous black curve represents the waveform obtained in the same conditions, though using for the instrument range PTR the Gaussian approximation Eq. (6). The dashed blue curve represents the waveform obtained with the square cardinal sine shape, this time neglecting the broadening induced by the surface wave motions (*ie* in the  $\tilde{\sigma}_f = \sigma_f$  approximation). The dashed black curve represents the waveform obtained with the Gaussian approximation of the PTR and in the  $\tilde{\sigma}_f = \sigma_f$  approximation. Finally, the red curve represent the multilook waveforms produced by the "pySAMOSA" [29] implementation of the sidelobe aliasing. The normalization is consistent with Figure 7.

The red curve in Figure 8 represents the waveform obtained using the modified multilook SAMOSA model in the same environmental conditions, together with the waveforms obtained with Eq. (20) using the sinc<sup>2</sup> (blue curve) and Gaussian (black curve) approximations of the instrument range PTR. The effect of the Gaussian approximation is visible with a slight ( $\sim 4\%$ ) increase in peak height, and a very slightly reduced "toe" with respect to the more exact squared cardinal sine representation. Though this effect is clearly visible, it is not sufficient to explain the discrepancy with the SAMOSA waveform.

The dashed black curve represents the waveform obtained in the Gaussian range PTR approximation (consistent with SAMOSA), but neglecting the Doppler broadening due to the scattering facets motion (also consistent with SAMOSA). The match is now clearly excellent.

This at the same time validates the IASCO waveform model Eq. (20), but also hints at a strong impact of the scattering facets motion on the SAR altimeter waveform. As the difference between the continuous and dashed blue curves shows, neglecting this effect likely leads to a strong overestimation of the waveform peak height, as well as a noticeable underestimation of the waveform "toe" extension.

As a side remark, we notice that the Gaussian approximation of the instrument range PTR, which is not practically useful as the final Fourier transform is performed numerically anyway, has actually a slight (supposedly detrimental) effect. We will henceforth not use it.

### C. Sensitivity to environmental conditions



Fig. 9. Graph of normalized altimeter waveforms for different 10-m neutral wind speeds, with significant wave height and vertical velocity variance computed from the spectrum [38] in its infinite-fetch limit. All waveforms are computed with the full DDA waveform Eq. (20), considering for the instrument range PTR the exact squared cardinal shape Eq. (5). The wind direction is orthogonal to the satellite track ( $\varepsilon = 0$ ). The influence of wind on signal level through variations of the backscattering cross-section  $\sigma_0$  is not accounted for.

1) Dependence on  $U_{10}$  at infinite fetch: The dependence of the IASCO waveform model Eq. (20) is represented as a function of 10-m neutral wind speed in Figure 9. The wind direction is orthogonal to the satellite track ( $\varepsilon = 0$ ). The wind acts through its influence on H<sub>s</sub> and  $\sigma_{\widetilde{w}}$ , which are computed using the spectrum prescription of [38] in its infinite-fetch limit. The dependence of the peak height on wind through the backscatter cross-section  $\sigma_0$  is not accounted for.

With increasing winds, the initial rise of the waveform progressively broadens, and the peak height progressively decreases. The point where the waveform is half-way up to its maximum height also progressively shifts to earlier instants. The decay part of the waveform is essentially impervious to parameter changes.



Fig. 10. Graph of normalized altimeter waveforms for fixed  $\sigma_{\widetilde{w}}=0.77~{\rm m.s^{-1}}$  and varying  ${\rm H_s}.$ 

2) Dependence on  $H_s$  for fixed  $\sigma_{\widetilde{w}}$ : Figure 9 mixes the influences of  $H_s$  and  $\sigma_{\widetilde{w}}$ , the two parameters affecting the waveform. Figure 10 shows the effect of changing the significant wave height for a fixed value of  $\sigma_{\widetilde{w}}$ , kept equal to its value in an infinite-fetch 12 m.s<sup>-1</sup>  $U_{10}$  sea surface, 0.77 m.s<sup>-1</sup>.  $\varepsilon$  is kept equal to 0.

With increasing  $H_s$ , the peak height decreases, and the initial rise progressively slackens. The half-height point gradually shifts to earlier instants.



Fig. 11. Graph of normalized altimeter waveforms for fixed  $\rm H_s=3.75~m$  and varying  $\sigma_{\widetilde{w}}.$ 

3) Dependence on  $\sigma_{\tilde{w}}$  for fixed H<sub>s</sub>: Figure 11 now conversely presents the effect of varying  $\sigma_{\tilde{w}}$  for a fixed H<sub>s</sub>, kept equal to its 12 m.s<sup>-1</sup> value of 3.75 m (refer to Figure 4 for

intuition on the range of variations explored here). Clearly, the influence of  $\sigma_{\widetilde{w}}$  on the waveform is quite different from the influence of  $H_s$ : with increasing  $\sigma_{\widetilde{w}}$ , the peak height decreases, the waveform "toe" lengthens, but this is achieved without shifting the half-height point of the peak.

One can see here a mechanism likely to bias the joint estimation of  $H_s$  and  $r_A$  from the waveform: in cases of particularly large  $\sigma_{\tilde{w}}$ , one might imagine the peak of the waveform to be quite low and the midpoint to be unaffected, which a tracker might mistake for a large  $H_s$  situation with a depressed sea surface.

4) Effect of the "Geophysical Doppler" shift ( $\varepsilon \neq 0$ ): The effect of a non-zero  $\varepsilon$  is extremely small, and would be very hard to appreciate from large-scale graphs such as Figures 9, 10 and 11. To provide insight into its influence on the waveform, the most practical approach is to study the biases its presence induces in the results of a waveform retracker such as SAMOSA. This is the subject of the next section.

### D. Tracker biases

The influences of  $U_{GD}$  and  $\sigma_{\tilde{w}}$  on the results of the SAMOSA retracker are studied in this section, by generating IASCO waveforms using known parameter values, and observing the biases of the retrieved values.



Fig. 12. Difference between SAMOSA retracked  $H_s$  and "true"  $H_s$  as a function of "true"  $H_s$ , obtained by varying  $U_{10}$  in the infinite-fetch limit of spectrum [38].

1) Significant height: Figure 12 presents the difference between "true"  $H_s$  and  $H_s$  estimates obtained running the SAMOSA retracker on IASCO waveforms generated with varying wind speeds. Again, the infinite-fetch form of the spectra is used to estimate  $H_s$ ,  $\sigma_{\tilde{w}}$  and  $U_{GD}$  as functions of  $U_{10}$ . Clearly, the produced waveforms are consistently associated by SAMOSA to overestimated values of  $H_s$ . This bias varies rapidly for small  $H_s$ , and increases weakly afterwards, rising from 0.22 m for  $H_s \simeq 2 m$  to 0.29 m for  $H_s \simeq 10.5 m$ . For a given wind speed, this bias is identical in the head-wind, tail-wind and cross-wind situations, showing it is essentially independent of  $\varepsilon$ , and dominated by the effect of  $\sigma_{\widetilde{w}}$ .



Fig. 13. SAMOSA retracked Sea Level Anomaly for a nominally flat surface located at  $\eta = 0$ , obtained by varying  $U_{10}$  in the infinite-fetch limit of spectrum [38].

2) Sea Level Anomaly: Figure 13 now presents the Sea Level Anomaly (SLA) estimates obtained using the SAMOSA retracker on IASCO waveforms generated for varying wind speeds blowing over a flat surface located at  $\eta = 0$ . Clearly, the biasing effect mentioned above is present: a quite large bias, varying between -1 cm and 2 cm is present even for zero  $\varepsilon$  (black curve). The mere existence of a non-zero  $\sigma_{\widetilde{w}}$  induces a bias in the retrieved SLA. This time, however, the effect of a non-zero "Geophysical Doppler" shift, though very small, is clearly visible. The existence of a head wind lures the tracker into overestimating the SLA, while the existence of a tail wind has the opposite effect. The azimuthal dependence of the bias (not shown) is essentially sinusoidal, but is not centered around 0, as the  $\sigma_{\widetilde{w}}$  effect introduces a (actually larger) isotropic bias.

### E. Effect of instrument parameters

The IASCO waveform can be applied with no modifications to the CryoSAT-2 and Sentinel-3A cases. However, as the altitude, flight velocity and  $f_p$  of these instruments are all different, it is hard to disentangle from raw waveform plots the changes due to the various parameters. This section is thus devoted to a discussion of the sensitivity of the waveform to the most important parameters, the pulse repetition frequency and the platform flight altitude. The two other main parameters are the antenna radiation diagram beamwidth and the flight velocity. The antenna beamwidth affects the waveform in a straightforward way, by changing the slope of the decay region of the waveform (the finer the beamwidth, the faster the decay). The flight velocity is not a very sensitive parameter over its quite limited range of variation, and mainly affects the balance between the main lobe and sidelobes of the along-track SAR processing, by changing the portion of the instrument FoV that is associated to Doppler shifts smaller than the Nyquist frequency.



Fig. 14. Graph of altimeter waveforms obtained for a  $12 \text{m.s}^{-1}$  side wind, corresponding to Hs = 3.75 m,  $\sigma_{\tilde{w}} = 0.77 \text{ m.s}^{-1}$ ,  $\varepsilon = 0$ , for the S6-MF flight parameters and for instrument  $f_p$  equal to varying fractions of the actual S6-MF  $f_p$ .

1) Effect of instrument  $f_p$ : Figure 14 shows the dependence of the IASCO waveform as a function of the instrument  $f_p$ , expressed in fraction of the actual S6-MF  $f_p$ , for constant burst duration (i.e. the number of pulses per burst is increased when  $f_p$  increases). Reducing the instrument  $f_p$  by a factor of two (cyan curve) reduces the effectiveness of the RMC, and reduces the main peak height by a large factor with respect to the nominal (orange) curve. Interestingly, increasing the  $f_p$  by a factor of two, thus bringing it very close to the CryoSAT-2 and Sentinel-3A  $f_p$  (green curve), seems to almost saturate this effect. Further increasing  $f_p$  only brings minimal changes (the black curve represents the  $f_p \rightarrow \infty$  limit). This is in fact not unexpected, as for large values of  $f_p$  the ambiguous frequency ranges correspond to very distant along-track portions of the sea surface, whose signal is effectively rejected by the antenna radiation diagram.

2) Effect of instrument  $N_p$ : Conversely, Figure 15 shows the dependence of the IASCO waveform on the burst duration (number of pulses per burst), for fixed  $f_p$ . The burst duration influences the waveform through the along-track SAR processing resolution  $\sigma_f$ . Changing  $N_b$  changes the measurement floor on  $\sigma_{\tilde{w}}$ , shown as the red dashed line in Figure 4. Higher  $N_b$  values reduce the instrumental contribution to the overall  $\widetilde{\sigma_f}$ . As can be seen in the figure, though  $N_b = 64$ , the value implemented in the S6-MF onboard processor, is a bit marginal, the WF obtained for a value of  $N_b = 128$  (green curve) is very close to the ideal curve obtained in the limit of vanishing instrumental contribution (black curve) in these conditions. Going much further could increase the influence of the range walk of scatters during the burst duration. Also, the finite lifetime of the individual scatters, which is currently not well known, sets limits to what can be achieved by increasing the coherent processing time window. Reprocessing S6-MF data with  $N_b = 128$  is feasible on ground, but can not be done using the current onboard processor.



Fig. 15. Graph of altimeter waveforms obtained for a  $12 \text{m.s}^{-1}$  side wind, corresponding to Hs = 3.75 m,  $\sigma_{\tilde{w}} = 0.77 \text{ m.s}^{-1}$ ,  $\varepsilon = 0$ , for the S6-MF flight parameters and  $f_p$ , and for for varying instrument  $N_b$  (burst duration).



Fig. 16. Graph of altimeter waveforms obtained for a  $12 \text{m.s}^{-1}$  side wind, corresponding to Hs = 3.75 m,  $\varepsilon = 0$ , for varying platform altitude, the S6-MF  $f_p$  and constant 7200 m.s<sup>-1</sup> platform speed. A compensation factor of  $(h/h_0)^{5/2}$  has been applied.

3) Effect of platform altitude: Figure 16 shows the dependence of the IASCO waveform on the platform altitude h, compensated for the expected  $h^{-5/2}$  power decrease [13]. The remaining dependence after amplitude compensation is small, and is essentially due to changes in energy balance between the main lobe and the sidelobes of the along-track SAR processing. It almost disappears in the  $\tilde{\sigma}_f \rightarrow 0$  limit. The only difference remains in the decay region. It is consistent with the change in the aliasing lobe slope in the DDM caused by the change of  $\mu$ . In this plot, the pulse repetition frequency has been kept fixed and equal to its S6-MF value.

### VI. CONCLUSION

With the ambition to more precisely estimate the impact of ocean surface motions on Delay-Doppler altimetry, a new analytical (Fourier-space in range) Delay Doppler Altimeter waveform has been derived. Developments start with the DDM signature of an isolated backscattering facet of the sea surface, the basic building block from which the aggregated DDM response of the sea surface is composed. Through a careful analysis of the joint statistics of its equivalent radar crosssection, its elevation with respect to the mean sea level, and its instantaneous velocity with respect to the radar instrument, the ensemble average of this signature is obtained under the Gaussian statistical approximation.

The convolution of this signature with the Flat Surface Impulse Response is computed, yielding an analytical expression of the full DDM, Eq. (23). Integrating this expression in Doppler, the IASCO waveform, an analytical expression of the DDA waveform, is finally obtained, Eq. (20). This waveform is validated against the well-established SAMOSA model waveform, modified to account for the SAR aliasing sidelobes.

Analytical, IASCO waveforms and related sensitivities can be readily obtained with respect to technological and environmental parameters. In particular, the influence of surface facets orbital and mean motion on the retracking process has been studied, using the SAMOSA retracker as a benchmark. Surface phenomena not taken into account by the SAMOSA waveform model, but addressed by the IASCO waveform model, are found to potentially cause observable biases in geophysical parameters of crucial scientific importance, such as surface waves significant height and Sea Level Anomaly.

This opens a number of potential options to circumvent, or take advantage of, these issues:

- On the one hand, a first, straightforward, option might be to combine the SAMOSA and IASCO waveforms in a retracker to retrieve and compare directly, through a Maximum-Likelihood approach, estimates of surface waves vertical motion standard deviation  $\sigma_{\tilde{w}}$  and of the along-track projection of the "Geophysical Doppler" vector U<sub>GD</sub>.
- These new parameters should find scientific use in their own right, or be used in parameterizations to input altimeter ESB- or SSB-mitigation routines.
- An hybrid approach might also be attempted, using CA estimates, which are impervious to the motion-induced biases DDA studied here, as first guesses or constraints for the DDA retracker. It will lead to the implementation of a Maximum A Posteriori optimization approach ingesting both the DDA and CA instrument waveforms.
- Another possible option would be to apply the fitting procedure to the DDM itself. Our Eq. (23) could serve as the basis for such an attempt.

Future investigations will be conducted to further dwell on these theoretical and analytical means to analyze actual and future 2D off-nadir altimeter performances. Efforts will also be conducted to optimize and extend processing strategies (e.g. [39]) to retrieve upper ocean sea surface height slopes and motions, to help extract upper ocean velocities.

$$\begin{array}{l} \text{Appendix A} \\ \text{Expressing } P(\partial_x \tilde{z} = \partial_x \tilde{z}|_{sp}, \partial_y \tilde{z} = \partial_y \tilde{z}|_{sp}, \tilde{w}). \end{array}$$

From section 2.1 in [35], an expression can be simply derived for  $P(\partial_x \tilde{z} = \partial_x \tilde{z}|_{sp}, \partial_y \tilde{z} = \partial_y \tilde{z}|_{sp}, \tilde{w})$  in terms of spectral moments of the sea surface elevation. In the notations of [3], [40],

$$m_{tt} = \left\langle \partial_t \tilde{z} \partial_t \tilde{z} \right\rangle, \ \mathbf{msv} = \left| \begin{array}{c} \left\langle \partial_t \tilde{z} \partial_x \tilde{z} \right\rangle \\ \left\langle \partial_t \tilde{z} \partial_y \tilde{z} \right\rangle \end{array} \right\rangle$$
$$\mathbf{M_{ss}} = \left[ \begin{array}{c} \left\langle \partial_x \tilde{z} \partial_x \tilde{z} \right\rangle & \left\langle \partial_x \tilde{z} \partial_y \tilde{z} \right\rangle \\ \left\langle \partial_x \tilde{z} \partial_y \tilde{z} \right\rangle & \left\langle \partial_y \tilde{z} \partial_y \tilde{z} \right\rangle \end{array} \right]$$

the correlation matrix of surface vertical velocity and surface slope components is:

$$\mathbf{M}_{\mathbf{tss}} = \begin{bmatrix} m_{tt} & \mathbf{msv}^T \\ \hline \mathbf{msv} & \mathbf{Mss} \end{bmatrix}$$

and their joint pdf :

$$P(\partial_x \tilde{z}, \partial_y \tilde{z}, \tilde{w}) = \frac{1}{(2\pi)^{3/2} \sqrt{|\det(\mathbf{M}_{tss})|}} \times \exp\left(-\frac{1}{2} \left[\tilde{w}, \partial_x \tilde{z}, \partial_y \tilde{z}\right] \mathbf{M}_{tss}^{-1} \middle| \begin{array}{c} \tilde{w} \\ \partial_x \tilde{z} \\ \partial_y \tilde{z} \end{array} \right).$$

The pdf of  $\tilde{w}$  at  $\partial_x \tilde{z} = \partial_x \tilde{z}|_{sp}, \partial_y \tilde{z} = \partial_y \tilde{z}|_{sp}$  is obtained by rearranging this expression. Using the Sherman-Morrison-Woodbury matrix inversion formula, the block inverse of  $\mathbf{M}_{tss}$ reads:

$$\mathbf{M_{tss}}^{-1} = \frac{1}{\sigma_{\widetilde{w}}^2} \begin{bmatrix} 1 & | -\mathbf{M_{ss}}^T \mathbf{M_{ss}}^{-1} \\ \hline -\mathbf{M_{ss}}^{-1} \mathbf{msv} & \mathbf{C_{ss}} \end{bmatrix},$$

with

$$\sigma_{\widetilde{w}}^{2} = m_{tt} - \mathbf{msv}^{T} \mathbf{M_{ss}}^{-1} \mathbf{msv},$$
  
$$\mathbf{C_{ss}} = \sigma_{\widetilde{w}}^{2} \mathbf{M_{ss}}^{-1} + \mathbf{M_{ss}}^{-1} \mathbf{msv} \mathbf{msv}^{T} \mathbf{M_{ss}}^{-1}$$

Also:

$$\det(\mathbf{M}_{\mathbf{tss}}) = \sigma_{\widetilde{w}}^2 \det(\mathbf{M}_{\mathbf{ss}})$$

One recognizes in  $-\mathbf{M_{ss}}^{-1}$  msv the "Geophysical Doppler" vector  $\mathbf{U_{GD}}$  [3]. This vector is composed of the Total Surface Current Vector and a (usually dominant) "Wave Doppler" vector. The magnitude of  $\mathbf{U_{GD}}$  and the azimuth towards which it points are respectively denoted as  $U_{GD}$  and  $\varphi_{GD}$ . Figure 17 represents  $U_{GD}$ ,  $\sqrt{m_{tt}}$  and  $\sigma_{\widetilde{w}}$  as a function of wind speed, using the elevation spectrum model of [38] for infinite fetch, and under the assumption of infinitely short electromagnetic wavelength (see discussion at the end of this section). Already discussed in [3], [40],  $U_{GD}$  varies rapidly with wind speed up to  $|U_{10}| \sim 7 \text{ m.s}^{-1}$ , after which it levels off on a slowly increasing plateau. The other quantities,  $\sqrt{m_{tt}}$  and  $\sigma_{\widetilde{w}}$ , increase almost monotonously,  $\sqrt{m_{tt}}$  remaining slightly larger than  $\sigma_{\widetilde{w}}$ .



Fig. 17. Spectral moments of  $\partial_t \eta$  as a function of wind speed, for the elevation spectrum model [38] at infinite fetch, expressed in m.s<sup>-1</sup> (left axis) and in Hz (right axis).

The pdf of  $\tilde{w}$  for a given surface slope  $\nabla \tilde{z}|_{sp}$  is:

$$P(\tilde{w}, \nabla \tilde{z}_{sp}) = \frac{1}{(2\pi)^{3/2} \sigma_{\tilde{w}} \sqrt{|\det(\mathbf{M}_{ss})|}} \times \exp\left(-\frac{1}{2\sigma_{\tilde{w}}^2} \left[\tilde{w} + \mathbf{U}_{\mathbf{GD}} \cdot \nabla \tilde{z}|_{sp}\right]^2\right) \times \exp\left(-\frac{1}{2} \nabla \tilde{z}|_{sp}^T \mathbf{M}_{ss}^{-1} \nabla \tilde{z}|_{sp}\right)$$

The term on the first line is a normalization factor, the term on the second line shows that the mode of the distribution of  $\tilde{w}$  is displaced by the projection of  $-\mathbf{U}_{GD}$  along the radar line-of-sight, and the final term accounts for the decreasing probability of finding specular facets away from nadir. The surface elevation gradient  $\nabla \tilde{z}|_{sp}$  on specular facets at the current point is fixed by the observation geometry, and such that  $\partial_x \eta|_{sp} = \tan(\psi) \sin(\varphi)$ , and  $\partial_y \eta|_{sp} = \tan(\psi) \cos(\varphi)$ . The  $\mathbf{U}_{GD} \cdot \nabla \tilde{z}|_{sp}$  term thus reduces to  $U_{GD} \tan(\psi) \cos(\varphi - \varphi_{GD})$ . The surface slope covariance matrix  $\mathbf{M}_{ss}$ , being symmetric, can be diagonalized by a mere coordinate system rotation, and a maximal surface slope variance azimuth  $\varphi_{ss}$  can be found such that:

$$\mathbf{M}_{\mathbf{ss}} = \mathbf{R}_{\varphi_{\mathbf{ss}}} \times \begin{bmatrix} \mathbf{m}_{\mathbf{a}} & \mathbf{0} \\ \mathbf{0} & \mathbf{m}_{\mathbf{x}} \end{bmatrix} \times \mathbf{R}_{-\varphi_{\mathbf{ss}}},$$

with  $\mathbf{R}_{\varphi_{ss}}$  the matrix describing rotation of angle  $\varphi_{ss}$  in the (x, y) plane, and  $m_a$  and  $m_x$  the mean squared surface slopes in the "along" and "across" directions. With these notations, one obtains

$$\begin{split} P(\tilde{w}, \nabla \tilde{z}_{sp}) &= \frac{1}{(2\pi)^{3/2} \sigma_{\tilde{w}} \sqrt{\mathrm{m_a m_x}}} \\ &\times \exp\left(-\frac{1}{2\sigma_{\tilde{w}}^2} \left[\tilde{w} + U_{GD} \tan(\psi) \cos(\varphi - \varphi_{GD})\right]^2\right) \\ &\times \exp\left(-\frac{\tan^2 \psi}{2} \left[\frac{\cos^2(\varphi - \varphi_{ss})}{\mathrm{m_a}} + \frac{\sin^2(\varphi - \varphi_{ss})}{\mathrm{m_x}}\right]\right). \end{split}$$

As in [12], we denote the total mean squared surface slope by  $mss_T = m_a + m_x$ . The fractional difference between  $m_a$  and

 $m_x$  is described by  $\delta_{ss}$ , such that  $m_a = \frac{mss_T}{2}(1 + \delta_{ss})$  and  $m_x = \frac{mss_T}{2}(1 - \delta_{ss})$ . With these notations

$$P(\tilde{w}, \nabla \tilde{z}_{sp}) = \frac{1}{\pi \sqrt{2\pi} \sigma_{\tilde{w}} \mathrm{mss}_{\mathrm{T}} \sqrt{1 - \delta_{ss}^2}} \\ \times \exp\left(-\frac{1}{2\sigma_{\tilde{w}}^2} \left[\tilde{w} + U_{GD} \tan(\psi) \cos(\varphi - \varphi_{GD})\right]^2\right) \\ \times \exp\left(-\frac{\tan^2 \psi}{\mathrm{mss}_{\mathrm{T}}(1 - \delta_{ss}^2)} \left[1 - \delta_{ss} \cos(2(\varphi - \varphi_{ss}))\right]\right).$$

The integrals yielding the mean squared slope statistics as moments of the surface elevation spectrum logarithmically diverge with wavevector k: they are thus dependent on the structure of the sea surface at all scales. This ultraviolet divergence can be regularized either by the roll-off of the elevation spectrum in the viscous regime, or by the filtering effect caused by the finiteness of the electromagnetic wavelength, depending on which limit is reached first. Numerous studies introduced mss<sub>shape</sub> to account for the electromagnetic filtering effect.

In the optical limit, [41]–[43] show  $m_x/m_a$  to be weakly dependent on environmental conditions and close to 0.6. In the present notations, this translates to  $\delta_{ss} \simeq 0.25$ .

Admittedly, the relevance of these optical results to Kuband physics remains to be firmly established, but point to a small impact of directional effects on the backscattering crosssection in the near-nadir geometry [44]. Taking the  $\delta_{ss} \rightarrow 0$ limit in the above expression:

$$P(\tilde{w}, \nabla \tilde{z}_{sp}) = \frac{1}{\pi \sqrt{2\pi} \sigma_{\tilde{w}} \text{mss}_{\text{T}}} \exp\left(-\frac{\tan^2 \psi}{\text{mss}_{\text{T}}}\right) \\ \times \exp\left(-\frac{1}{2\sigma_{\tilde{w}}^2} \left[\tilde{w} + U_{GD} \tan(\psi) \cos(\varphi - \varphi_{GD})\right]^2\right).$$

Note one should probably replace  $mss_T$  by  $mss_{shape}$  to empirically account for the electromagnetic cut-off effect. In our calculations, integrations are performed up to the viscous cut-off limit, using the spectral form proposed by [38], to keep  $mss_T$  in the developments.

### APPENDIX B FROM FSIR TO ECHO WAVEFORM

### A. Fourier transformation of the FSIR

Starting from Eq. (15), the Fourier transform in r and f of the FSIR can be expressed as

$$\widehat{\text{FSIR}}(\mathcal{T},\mathcal{K}) = \mathcal{A}_{\varepsilon} e^{-i[\mathcal{T}f_A + \mathcal{K}r_A]} \\ \times \iint df dr e^{-i\mathcal{T}f - r(\nu + i\mathcal{K})} \frac{H(r - \mu_{\varepsilon}f^2)}{\sqrt{r - \mu_{\varepsilon}f^2}}.$$

Performing the change of variable  $\sqrt{r-\mu_{\varepsilon}f^2} \to u$  turns this expression into

$$\widehat{\widehat{\mathrm{FSIR}}}(\mathcal{T},\mathcal{K}) = \mathcal{A}_{\varepsilon} e^{-i[\mathcal{T}f_A + \mathcal{K}r_A]} \\ \times \int \mathrm{d}f e^{-i\mathcal{T}f - (\nu + i\mathcal{K})\mu_{\varepsilon}f^2} \int e^{-u^2(\nu + i\mathcal{K})} \mathrm{d}u.$$

The two Gaussian integrations can be performed independently. Introducing the approximations  $\mu_{\varepsilon} \simeq \mu_0(1-2\varepsilon)$  and  $\mathcal{A}_{\varepsilon} \simeq \mathcal{A}_0(1-\varepsilon)$ , one obtains the final result as

$$\widehat{\widehat{\mathrm{FSIR}}}(\mathcal{T},\mathcal{K}) = \mathcal{A}_0 \frac{\pi e^{-i[\mathcal{T}f_A + \mathcal{K}r_A]} e^{-\frac{\mathcal{T}^2}{4\mu_\varepsilon(\nu + i\mathcal{K})}}}{\sqrt{\mu_0}(\nu + i\mathcal{K})}$$

### *B. Fourier transformation of the wave-accounting Point Target Response*

Starting from the expressions (9) and (12) of the waveaccounting PTRs, and using the fact that

$$\int \mathrm{d}u e^{-iu\mathcal{K}} e^{-\frac{u^2}{2\sigma^2}} = \sqrt{2\pi}\sigma e^{-\frac{\mathcal{K}^2\sigma^2}{2}},$$

the Fourier transform of the PTR is easily obtained as

$$\widehat{\widetilde{\widetilde{\Upsilon_f^2}}\widetilde{\Upsilon_r^2}}(\mathcal{T},\mathcal{K}) = \sqrt{2\pi}\sigma_f \frac{25^2}{46^2} e^{-\frac{\tau^2 \overline{\sigma_f}^2 + \mathcal{K}^2 \sigma_h^2}{2}} \widehat{\Upsilon_r^2}(\mathcal{K}).$$

Depending on the intended use of the expression, either the exact representation of the instrument PTR Eq. (5) or its Gaussian approximation Eq. (6) can be used for  $\widehat{\Upsilon_r^2}(\mathcal{K})$ .

### C. Computation of the CA echo waveform from Equation (16)

Going from expression (16) to the DDA waveform still requires significant work. The succession of steps is detailed in the main text: the Fourier-space DDM must still be inversetransformed in the Doppler dimension, the RMC must be applied, the result must be stacked by integration in the Doppler dimension, and the result must be inverse-transformed in the range dimension to yield the waveform.

In the CA context, however, no RMC is applied before the integration. The range Fourier-transform of the CA waveform is equal to the Doppler-dimension integral of the DDM, which is easily obtained as the value of Eq. (16) along the T = 0 line:

$$\widehat{W_{CA}}(\mathcal{K}) = \mathcal{A}_0 \frac{\sqrt{2\pi}\pi\sigma_f}{\sqrt{\mu_0}} \frac{25^2}{46^2} e^{-i\mathcal{K}r_A} \frac{e^{-\frac{\mathcal{K}^2\sigma_h^2}{2}}}{\nu + i\mathcal{K}} \widehat{\Upsilon}_r^2(\mathcal{K}).$$

Assuming the Gaussian shape for  $\hat{\Upsilon}_r^2$ , the inverse Fourier transform of this function can be obtained using Eq. 3.462.3 of [45] in terms of the parabolic cylinder function  $D_{-1}$ . This function can in turn be expressed in terms of the complementary error function, and the result is obtained as:

$$W_{CA}(r) = \mathcal{A}_0 \frac{\pi^2 \sigma_r \sigma_f}{\sqrt{\mu_0}} \frac{25^2}{46^2} e^{-\frac{\nu^2 \widetilde{\sigma_r}^2}{2}}$$
(21)  
 
$$\times \exp\left(-\nu \left[r - r_A - \nu \widetilde{\sigma_r}^2\right]\right) \operatorname{erfc}\left(-\frac{r - r_A - \nu \widetilde{\sigma_r}^2}{\sqrt{2}\widetilde{\sigma_r}}\right).$$

This expression coincides with the classical result [1], [22]. The expression of  $r_A$  contains a  $\frac{\delta t_{rr}^2 v_t^2}{2\kappa h}$  shift, which amounts to 0.146 mm: due to the range-Doppler ambiguity bias, the sea surface appears 0.146 mm closer to the satellite than it actually is. The effective interface thickness  $\tilde{\sigma_r}$ , accounting for both the instrument PTR and the sea surface height dispersion due to waves, has its usual expression.

### D. Range Migration Correction and stacking

In this step, the expression of  $\widehat{DDM}(f, \mathcal{K})$  given by Eq. (17) has to be multiplied by  $\exp(i\mu_0\mathcal{K}f^2)$ , and the result must be integrated ("stacked") in f. The product of the DDM Fourier transform and the RMC mask is:

$$\widehat{\text{DDM}}_{\text{RMC}}(f,\mathcal{K}) = \mathcal{A}_{\varepsilon} \sqrt{2\pi} \sigma_f \frac{25^2}{46^2} \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}}}{\sqrt{\nu + i\mathcal{K}}} \widehat{\Upsilon}_r^2(\mathcal{K}) \\ \times \frac{e^{-f^2 \left[\frac{\mu_{\varepsilon}(\nu + i\mathcal{K})}{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\overline{\sigma_f}^2} - i\mathcal{K}\mu_0\right]}}{\sqrt{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\overline{\sigma_f}^2}}.$$

Introducing  $\Xi_{\varepsilon}$  such that

$$\Xi_{\varepsilon}^{2} = \frac{\mu_{\varepsilon}(\nu + i\mathcal{K})}{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma_{f}}^{2}} - i\mathcal{K}\mu_{0}$$

it can be expressed as

$$\widehat{\text{DDM}}_{\text{RMC}}(f,\mathcal{K}) = \mathcal{A}_{\varepsilon} \sqrt{2\pi} \sigma_{f} \frac{25^{2}}{46^{2}} \frac{e^{-i\mathcal{K}r_{A} - \frac{\mathcal{K}^{2}\sigma_{h}^{2}}{2}}}{\sqrt{\nu + i\mathcal{K}}} \widehat{\Upsilon}_{r}^{2}(\mathcal{K}) \\ \times \frac{e^{-\Xi_{\varepsilon}^{2}f^{2}}}{\sqrt{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma_{f}}^{2}}}.$$
 (22)

Integrating this expression in f yields the Fourier-transform in r of the DDA waveform. The integral is Gaussian and is easily performed, and the result is obtained as

$$\begin{split} \widehat{W_{DDA}}(\mathcal{K}) &= \mathcal{A}_{\varepsilon} \sqrt{2} \pi^{3/2} \sigma_f \frac{25^2}{46^2} \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}}}{\sqrt{\nu + i\mathcal{K}}} \widehat{\Upsilon_r^2}(\mathcal{K}) \\ &\times \frac{1}{\Xi_{\varepsilon} \sqrt{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma_f}^2}}. \end{split}$$

This expression now deserves being simplified. Using again the approximations  $\mu_{\varepsilon} \simeq \mu_0(1-2\varepsilon)$  and  $\mathcal{A}_{\varepsilon} \simeq \mathcal{A}_0(1-\varepsilon)$ , and neglecting terms of order  $\varepsilon^2$ , one obtains

$$\Xi_{\varepsilon} \simeq \sqrt{\mu_{\varepsilon} \frac{\nu - 2i\mathcal{K}(\varepsilon + \mu_{0}\nu\widetilde{\sigma_{f}}^{2}) + 2\mu_{0}\mathcal{K}^{2}\widetilde{\sigma_{f}}^{2}}{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma_{f}}^{2}}}$$

and

$$\widehat{W_{DDA}}(\mathcal{K}) = \mathcal{A}_0 \sqrt{2} \pi^{3/2} \sigma_f \frac{25^2}{46^2} \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}}}{\sqrt{\mu_0}\sqrt{\nu + i\mathcal{K}}} \widehat{\Upsilon}_r^2(\mathcal{K}) \\ \times \frac{1}{\sqrt{\nu - 2i\mathcal{K}\left[\varepsilon + \mu_0 \nu \widetilde{\sigma_f}^2\right] + 2\mu_0 \mathcal{K}^2 \widetilde{\sigma_f}^2}}.$$

### *E.* Analytical inversion in the $\varepsilon = 0$ , $\widetilde{\sigma_f} = 0$ case

Inverting this Fourier transform in the general case is difficult. Assuming a Gaussian range PTR, the special case  $\varepsilon = 0$ ,  $\widetilde{\sigma_f} = 0$  studied by [22] is however tractable, and provides a check of the expression. Starting from

$$W_{DDA}(r) = \mathcal{A}_0 \frac{\pi \sigma_r \sigma_f}{\sqrt{\mu_0 \nu}} \frac{25^2}{46^2} \int d\mathcal{K} \frac{e^{i\mathcal{K}(r-r_A)} e^{-\frac{\mathcal{K}^2 \widetilde{\sigma_r}^2}{2}}}{\sqrt{\nu + i\mathcal{K}}},$$

introducing the integration variable shift  $\mathcal{K} \to \mathcal{K} + i\nu$  brings the integral to the form

$$W_{DDA}(r) = \mathcal{A}_0 \frac{\pi \sigma_r \sigma_f}{\sqrt{\nu \mu_0}} \frac{25^2}{46^2} e^{-\nu(r-r_A - \nu \frac{\widetilde{\sigma_r}^2}{2})} \\ \int (i\mathcal{K})^{-1/2} e^{i\mathcal{K}(r-r_A - \nu \widetilde{\sigma_r}^2)} e^{-\frac{\mathcal{K}^2 \widetilde{\sigma_r}^2}{2}} d\mathcal{K}.$$

Using Eq. 3.462.3 of [45] then yields

$$W_{DDA}(r) = \mathcal{A}_0 \sigma_r \sigma_f \frac{25^2}{46^2} \frac{\sqrt{2}\pi^{3/2}}{\sqrt{\widetilde{\sigma_r}\nu\mu_0}} \mathcal{D}_{-1/2} \left( -\frac{r - r_A - \nu\widetilde{\sigma_r}^2}{\widetilde{\sigma_r}} \right) \\ \exp\left( -\frac{\nu^2 \widetilde{\sigma_r}^2}{2} - \nu(r - r_A - \nu\widetilde{\sigma_r}^2) - \frac{(r - r_A - \nu\widetilde{\sigma_r}^2)^2}{4\widetilde{\sigma_r}^2} \right)$$

where  $D_{-1/2}$  is the parabolic cylinder function of order -1/2.

Up to changes in the normalization constant and the transposition from r to two-way travel time, this expression is identical to Eq. (A16) of [22].

As noted by these authors, numerical implementations of this parabolic cylinder function have a tendency to overflow for negative argument. This problem can be circumvented using Eqs. (19.3.7, 19.3.8, 19.27.4 and 19.27.5) from [46] to express it in terms of modified Bessel functions as:

$$D_{-1/2}(x) = \sqrt{\frac{|x|}{2\pi}} K_{1/4}\left(\frac{x^2}{4}\right) + H(-x)\sqrt{\pi|x|} I_{1/4}\left(\frac{x^2}{4}\right).$$

A more uniform balancing of the different terms of the expression can then be achieved using the exponentially scaled implementations of the modified Bessel functions.

### F. Accounting for sidelobes in the DDM

For  $f \in [-f_p/2; f_p/2]$ , the full DDM, accounting for Doppler sidelobes, can be expressed as:

$$\begin{split} & \widehat{\text{DDM}}_{\text{RMC}}^{\text{SL}}(f,\mathcal{K}) = e^{i\mu_0\mathcal{K}f^2} \\ & \times \left[ \widehat{\text{DDM}}(f,\mathcal{K}) + \widehat{\text{DDM}}(f+f_p,\mathcal{K}) + \widehat{\text{DDM}}(f-f_p,\mathcal{K}) \right] \end{split}$$

Starting from Eq. (17), this is easily expressed as:

$$\widehat{\text{DDM}}_{\text{RMC}}^{\text{SL}}(f,\mathcal{K}) = \mathcal{A}_{\varepsilon}\sqrt{2}\pi\sigma_{f}\frac{25^{2}}{46^{2}}\widehat{\Upsilon}_{r}^{2}(\mathcal{K})$$
(23)  
 
$$\times \frac{e^{-i\mathcal{K}r_{A} - \frac{\mathcal{K}^{2}\sigma_{h}^{2}}{2}}e^{-\Xi_{\varepsilon}^{2}f^{2}}}{\sqrt{\nu + i\mathcal{K}}\sqrt{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma}_{f}^{2}}}$$
$$\times \left[1 + e^{-(\Xi_{\varepsilon}^{2} + i\mu_{0}\mathcal{K})f_{p}(f_{p} + 2f)} + e^{-(\Xi_{\varepsilon}^{2} + i\mu_{0}\mathcal{K})f_{p}(f_{p} - 2f)}\right].$$

### G. Accounting for sidelobes in the DDA waveform

Starting back from Eqs. (19) and (17), the main lobe contribution to the sidelobes-accounting DDA waveform can be expressed as

$$2\int_{0}^{\frac{f_{p}}{2}} e^{i\mu_{0}\mathcal{K}f^{2}}\widehat{\mathrm{DDM}}(f,\mathcal{K})\mathrm{d}f = 2^{3/2}\mathcal{A}_{\varepsilon}\pi\sigma_{f}\frac{25^{2}}{46^{2}}\widehat{\Upsilon}_{r}^{2}(\mathcal{K})$$
$$\times \frac{e^{-i\mathcal{K}r_{A}-\frac{\mathcal{K}^{2}\sigma_{h}^{2}}{2}}}{\sqrt{\nu+i\mathcal{K}}\sqrt{1+2\mu_{\varepsilon}(\nu+i\mathcal{K})\widetilde{\sigma_{f}}^{2}}}\int_{0}^{f_{p}/2} e^{-\Xi_{\varepsilon}^{2}f^{2}}\mathrm{d}f.$$

The integral can be expressed in terms of the error function. After simplification, this leads to

$$2\int_{0}^{\frac{f_{p}}{2}} e^{i\mu_{0}\mathcal{K}f^{2}} \widehat{\mathrm{DDM}}(f,\mathcal{K}) \mathrm{d}f = \mathcal{A}_{0} \frac{\sqrt{2}\pi^{3/2}}{\sqrt{\mu_{0}}} \sigma_{f} \frac{25^{2}}{46^{2}}$$
$$\times \frac{\widehat{\Upsilon}_{r}^{2}(\mathcal{K})e^{-i\mathcal{K}r_{A}} - \frac{\mathcal{K}^{2}\sigma_{h}^{2}}{2}\mathrm{erf}\left(\frac{f_{p}}{2}\Xi_{\varepsilon}\right)}{\sqrt{\nu + i\mathcal{K}}\sqrt{\nu - 2i\mathcal{K}(\varepsilon + \mu_{0}\nu\widetilde{\sigma_{f}}^{2}) + 2\mu_{0}\mathcal{K}^{2}\widetilde{\sigma_{f}}^{2}}}.$$

The contribution of the sidelobes, on the other hand, is

$$2\int_{\frac{f_p}{2}}^{\frac{3f_p}{2}} e^{i\mu_0\mathcal{K}(f-f_p)^2} \widehat{\text{DDM}}(f,\mathcal{K}) \mathrm{d}f = 2^{3/2}\mathcal{A}_{\varepsilon}\pi\sigma_f \frac{25^2}{46^2} \\ \times \frac{\widehat{\Upsilon_r^2}(\mathcal{K})e^{-\frac{\mathcal{K}^2\sigma_h^2}{2} + i\mathcal{K}[\mu_0f_p^2 - r_A]}}{\sqrt{\nu + i\mathcal{K}}\sqrt{1 + 2\mu_{\varepsilon}(\nu + i\mathcal{K})\widetilde{\sigma_f}^2}} \\ \times \int_{\frac{f_p}{2}}^{\frac{3f_p}{2}} e^{-\Xi_{\varepsilon}^2f^2 - i2\mu_0\mathcal{K}ff_p} \mathrm{d}f.$$

The integral can again be expressed in terms of the complementary error function erfc. The calculation is tedious, but straightforward, and the result is obtained as:

$$\begin{split} & 2\int_{\frac{f_p}{2}}^{\frac{3f_p}{2}} e^{i\mu_0\mathcal{K}(f-f_p)^2}\widehat{\mathrm{DDM}}(f,\mathcal{K})\mathrm{d}f = \mathcal{A}_0\frac{\sqrt{2}\pi^{3/2}}{\sqrt{\mu_0}}\sigma_f\frac{25^2}{46^2} \\ & \times \frac{\widehat{\Upsilon}_r^2(\mathcal{K})e^{-\mathcal{K}^2\frac{\sigma_h^2}{2}+i\mathcal{K}[\mu_0f_p^2-r_A]-\mathcal{K}^2\frac{\mu_0^2f_p^2}{\Xi_\varepsilon^2}}}{\sqrt{\nu+i\mathcal{K}}\sqrt{\nu-2i\mathcal{K}(\varepsilon+\mu_0\nu\widetilde{\sigma_f}^2)+2\mu_0\mathcal{K}^2\widetilde{\sigma_f}^2}} \\ & \times \left[\mathrm{erfc}\left(f_p\left[\frac{\Xi_\varepsilon}{2}+\frac{i\mu_0\mathcal{K}}{\Xi_\varepsilon}\right]\right)-\mathrm{erfc}\left(f_p\left[\frac{3\Xi_\varepsilon}{2}+\frac{i\mu_0\mathcal{K}}{\Xi_\varepsilon}\right]\right)\right]. \end{split}$$

Finally, the complete Fourier space expression of the sidelobes-accounting DDA waveform is obtained as

$$\begin{split} \widehat{W_{DDA}^{SL}}(\mathcal{K}) &= \mathcal{A}_0 \frac{\sqrt{2}\pi^{3/2}}{\sqrt{\mu_0}} \sigma_f \frac{25^2}{46^2} \widehat{\Upsilon_r^2}(\mathcal{K}) \\ &\times \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}}}{\sqrt{\nu + i\mathcal{K}} \sqrt{\nu - 2i\mathcal{K}(\varepsilon + \mu_0 \nu \widetilde{\sigma_f}^2) + 2\mu_0 \mathcal{K}^2 \widetilde{\sigma_f}^2}} \\ &\times \left[ \operatorname{erf}\left(\frac{f_p}{2} \frac{\Xi_{\varepsilon}}{2}\right) \\ &+ e^{i\mathcal{K}\mu_0 f_p^2 - \mathcal{K}^2 \frac{\mu_0^2 f_p^2}{\Xi_{\varepsilon}^2}} \operatorname{erfc}\left(f_p\left[\frac{\Xi_{\varepsilon}}{2} + \frac{i\mu_0 \mathcal{K}}{\Xi_{\varepsilon}}\right]\right) \\ &- e^{i\mathcal{K}\mu_0 f_p^2 - \mathcal{K}^2 \frac{\mu_0^2 f_p^2}{\Xi_{\varepsilon}^2}} \operatorname{erfc}\left(f_p\left[\frac{3\Xi_{\varepsilon}}{2} + \frac{i\mu_0 \mathcal{K}}{\Xi_{\varepsilon}}\right]\right) \right] \end{split}$$

Again, numerical implementations of the complementary error function have a tendency to overflow, and one has to resort to exponentially scaled implementations. Denoting  $\overline{\operatorname{erfc}}(x) = e^{x^2} \operatorname{erfc}(x)$ , a more stable expression is obtained as

$$\begin{split} \widehat{W_{DDA}^{SL}}(\mathcal{K}) &= \mathcal{A}_0 \frac{\sqrt{2}\pi^{3/2}}{\sqrt{\mu_0}} \sigma_f \frac{25^2}{46^2} \widehat{\Upsilon_r^2}(\mathcal{K}) \\ &\times \frac{e^{-i\mathcal{K}r_A - \frac{\mathcal{K}^2 \sigma_h^2}{2}}}{\sqrt{\nu + i\mathcal{K}} \sqrt{\nu - 2i\mathcal{K}(\varepsilon + \mu_0 \nu \widetilde{\sigma_f}^2) + 2\mu_0 \mathcal{K}^2 \widetilde{\sigma_f}^2}} \\ &\times \left[ \operatorname{erf}\left(\frac{f_p}{2} \frac{\Xi_{\varepsilon}}{2}\right) \\ &+ e^{-f_p^2 \frac{\Xi_{\varepsilon}^2}{4}} \overline{\operatorname{erfc}}\left(f_p \left[\frac{\Xi_{\varepsilon}}{2} + \frac{i\mu_0 \mathcal{K}}{\Xi_{\varepsilon}}\right]\right) \\ &- e^{-f_p^2 \left[i2\mathcal{K}\mu_0 + \frac{9\Xi_{\varepsilon}^2}{4}\right]} \overline{\operatorname{erfc}}\left(f_p \left[\frac{3\Xi_{\varepsilon}}{2} + \frac{i\mu_0 \mathcal{K}}{\Xi_{\varepsilon}}\right]\right) \right]. \end{split}$$

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Louis Marié received the B.Eng. degree from Ecole Nationale Supérieure des Mines de Paris, France, in 1999, and the Ph.D. degree in physics (fluid dynamics) from the Université Denis Diderot, Paris, France, in 2003. He is a Research Scientist with the Laboratoire d'Océanographie Physique et Spatiale, Institut Français de Recherche pour l'Exploitation de la Mer, Plouzané, France. He is a field oceanographer specialized in shelf seas dynamics. He has experience in non-linear physics, dynamo theory, turbulence theory and electromagnetism.



**Frédéric Nouguier** received the Agrégation and M.S. degrees in applied physics from the Ecole Normale Supérieure de Cachan, Cachan, France, the M.S. degree in physical methods for remote sensing from the University of Paris-Diderot, Paris, France, and the Ph.D. degree in physics from Aix-Marseille University, Marseille, France, in 2009. He is currently a Research Scientist with the Laboratoire d'Océanographie Physique et Spatiale, Institut Francais de Recherche pour l'Exploitation de la

Mer, Plouzané, France. He has experience in applied mathematics, physical oceanography, and electromagnetic wave theory and its application to ocean remote sensing.



**Doug Vandemark** received the bachelor's degree in physics from the Hope College, New York, NY, USA, in 1986, the master's degree in electrical engineering from the Microwave Remote Sensing Laboratory, University of Massachusetts at Amherst, Amherst, MA, USA, in 1988, and the Ph.D. degree in Earth science from the University of New Hampshire, Durham, NH, USA, in 2005.,He was a Research Engineer and a Scientist with the Goddard Space Flight Center, National Aeronautics and Space Administration (NASA), Washington, DC, USA,

from 1990 to 2005. He is active in studies tied to numerous NASA and European Space Agency Earth observing satellites and serves as a Principal Investigator on NASA's Ocean Surface Topography, Ocean Surface Salinity, Ocean Vector Wind, and Surface Water and Ocean Topography Science Teams. His research interests include the study of ocean and atmosphere boundary layer interactions and how to remotely measure near-surface ocean properties using radar and radiometer systems., Dr. Vandemark has served as a Strategic Advisory Team Member for the Northeast Regional Association of Coastal Ocean Observing Systems, the NASA-Jet Propulsion Laboratory (JPL) Physical Oceanography Data Archive Center, and the U.S. He has served as the Chair and a member of the NASA Earth Science Senior Reviews. He has served as an Associate Editor for the IEEE Geoscience and Remote Sensing Transactions.



Fabrice Ardhuin graduated from Ecole Polytechnique, Palaiseau, France, in 1997, and the Ph.D. degree in oceanography from the U.S. Naval Postgraduate School, Monterey, CA, USA, in 2001. He has been involved in ocean waves and related topics, from microseisms to remote sensing, with a strong focus on wave-current and wave-ice interactions, first at the French Navy Hydrographic and Oceanographic Service, Brest, France, then Institut Francais de Recherche pour l'Exploitation de la Mer, Plouzané, France, and now CNRS, at the Laboratory

for Ocean Physics and Satellite remote sensing. He led the SKIM proposal from 2016 to 2019 and is a member of the SWOT Science Team, the SEASTAR Mission Advisory Group, a co-investigator for the ODYSEA mission proposal, and the lead scientist for the ESA Sea State Climate Change Initiative project.



**Bertrand Chapron** received the B.Eng. degree from the Institut National Polytechnique de Grenoble, Grenoble, France, in 1984, and the Ph.D. degree in physics (fluid mechanics) from the University of Aix-Marseille II, Marseille, France, in 1988. He spent three years as a Post-Doctoral Research Associate with the NASA Goddard Space Flight Center, Greenbelt, MD, USA, and the Wallops Flight Facility, Wallops Island, VA, USA. He is currently a Research Scientist with the Laboratoire d'Océanographie Physique et Spatiale, Institut Fran-

cais de Recherche pour l'Exploitation de la Mer, Plouzané, France. He has experience in applied mathematics, physical oceanography, and electromagnetic wave theory and its application to ocean remote sensing.